

## COMPONENTS

Correctly identifying electronic parts can be one of the most difficult tasks facing someone building or repairing electronic equipment. As components become even smaller to allow higher density circuit boards, it is increasingly difficult to distinguish the different types such as resistors, capacitors, inductors and subminiature fuses.

Knowing what markings to expect can be a big help when faced with a circuit board full of unfamiliar components.

The information presented here, combined with a basic understanding of components should enable you to correctly identify those most commonly used.

## PREFERRED VALUES

The system of preferred values, which is used for resistors, capacitors and inductors, was developed to provide a logical progression from one value to the next, where each value represents an increase by an approximately constant percentage. Depending on the tolerance of the particular components, there can be between 3 and 192 preferred values in each decade.

The more common series are shown in the tables below. Values given for each series are repeated in every decade.

### 3 PER DECADE (50% TOLERANCE)

10 22 47

### 12 PER DECADE (10% TOLERANCE)

10 12 15 18 22 27 33 39  
47 56 68 82

### 24 PER DECADE (5% TOLERANCE)

10 11 12 13 15 16 18 20  
22 24 27 30 33 36 39 43  
47 51 56 62 68 75 82 91

## DECIMAL MULTIPLIERS

Decimal multiplier prefixes are in common use to simplify and shorten the notations of quantities such as component values.

Capacitance, for example, is measured in Farads. But the Farad is far too large a unit to be of practical use in most cases. For convenience, we use sub-multiples to save a lot of figures. For example, instead of writing 0.00000000001 Farads, we write 1pF (1 picofarad).

The more common prefixes and the relationships to one another are as follows.

ABBREV.	PREFIX	MULTIPLY BY	OR
p	pico	0.00000000001	10 <sup>-12</sup>
n	nano	0.000000001	10 <sup>-9</sup>
u	micro	0.000001	10 <sup>-6</sup>
m	milli	0.001	10 <sup>-3</sup>
–	UNIT	1	10 <sup>0</sup>
k	kilo	1000	10 <sup>3</sup>
M	mega	1000000	10 <sup>6</sup>

### UNITS

1000 pico units	=	1 nano unit
1000 nano units	=	1 micro unit
1000 micro units	=	1 milli unit
1000 milli units	=	1 unit
1000 units	=	1 kilo unit
1000 kilo units	=	1 mega unit

## CIRCUIT NOTATION

Some circuits give component values as they are normally spoken – e.g. 4.7pF for 4.7 picofarads, 5.6nH for 5.6 nanohenries. Others replace the decimal point with the first letter of the sub-multiple e.g. 5n6 for a 5.6nF capacitor or a 5.6nH inductor. Similarly for resistors, 6k8 is the same as 6.8k ohms while 1R5 would mean 1.5 ohms.

## TOLERANCE

All components differ from their marked value by some amount. Tolerance specifies the maximum allowed deviation from the specified value. Tolerances are normally expressed as a percentage of the nominal value.

As an example, a component with a marked value of 100 and a tolerance of 5%, could actually be any value between 5% below the marked value (95), and 5% above the marked value (105).

## RESISTORS

Most resistors are so small that it is impractical to print their values on them using normal numeric characters. Instead, they are marked using a code of coloured bands.

Resistors made to tolerance of 5% and 10% are marked with 4 bands while higher precision types, such as 2%, 1% or better, may be marked with 5 bands to allow for an extra digit of precision.

### HOW TO READ 4-BAND CODES:

At one end of the resistor there will be a gold, silver or brown tolerance band. This band is usually spaced apart from the other three bands. Start with the band nearest to the other end. Its colour represents the first digit of the resistor's value, as shown in the colour code chart. The next band represents the second digit of the resistor's value. The third band represents the decimal multiplier, that is, the number of zeros that we have to put after the first two digits to arrive at the resistor's value. The final band gives us the tolerance of the resistor, silver for 10% types, gold for 5% types, brown for 1% types.

Let's take the resistor shown at the top of the colour chart as an example. Its first band is yellow, representing '4' and the second band is violet, representing '7'. The third band, the multiplier, is orange which tells us to add 3 zeros to the number

we already have. This is the same as multiplying it by 1,000. Thus the value of the resistor is 47,000—forty seven thousand ohms or 47kΩ.

Finally, the fourth band, being gold, indicates that the resistor has a 5% tolerance, that is, its actual value will be somewhere between 44,650 and 49,350 ohms.

Some special high-voltage resistors use a yellow tolerance band in lieu of gold. This is simply because the metal particles in the gold paint might compromise the resistor's voltage rating.

### WHAT THEY MEAN:

- Band one – first figure of value
- Band two – second figure of value
- Band three – number of zeros/multiplier
- Band four – tolerance



### TOLERANCE BAND COLOURS:

Brown 1%, red 2%, gold 5%, silver 10%, none 20%.

### READING 5-BAND RESISTORS:

Because the final band on these resistors is usually brown or red, it can be a bit more difficult to know

which end to start from. In most cases the first four bands are grouped a bit closer together than the fourth and fifth bands. The first two bands are read the same as they are on the 4-band types. The third band supplies the third digit of the value. The fourth band now becomes the multiplier and the fifth represents the tolerance.

For example, if the 5 bands are, from first to fifth, yellow/orange/black/red/brown, then the three significant digits of the value would be 430', the multiplier would be 2, and the tolerance 1%. Hence, this is the code for a 43000 ohm, 1% resistor.

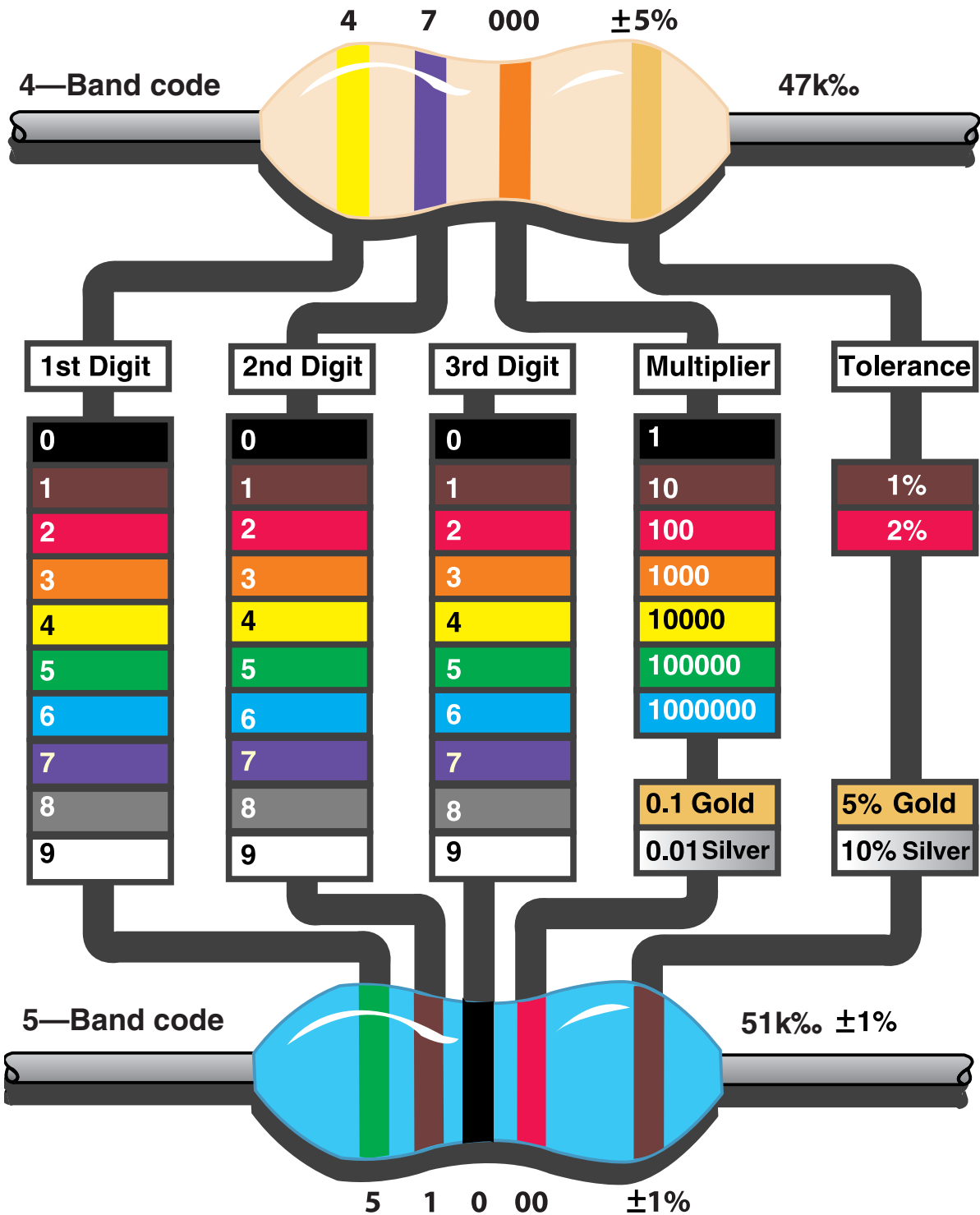
### WHAT THEY MEAN:

- Band one – first figure of value
- Band two – second figure of value
- Band three – third figure of value
- Band four – number of zeros/multiplier
- Band five – tolerance



RESISTOR COLOUR CODE

# RESISTOR COLOUR CODE



## CAPACITORS

Capacitors may be marked to show their value, voltage rating, accuracy, temperature stability and other information. Most capacitors are not marked with all of these, however, the value and voltage rating are usually given. Identification can be difficult because of the variety of systems in use.

### UNITS

The unit of capacitance is the Farad, but this unit is too large in practice. Commonly used smaller units are the microfarad (abbreviated  $\mu\text{F}$ ), nanofarad (nF) and picofarad (pF). The section on decimal multipliers (page 274) shows the relationship between these. Some capacitance values are commonly expressed by only one unit while others can be expressed under two or more units, e.g.  $1\mu\text{F}$  would rarely be called  $1000\text{nF}$  and never  $1,000,000\text{pF}$ , even though these are equivalent. However,  $0.0047\mu\text{F}$  is often expressed as  $4.7\text{nF}$ , or as  $4700\text{pF}$ .

### VALUE

Larger capacitors are marked in microfarads and indicate this by the abbreviations 'uF', 'u' or even the obsolete 'MFD'. Smaller capacitors are marked in nanofarads or picofarads and may abbreviate the unit to 'n' or 'p'.

If the value contains a decimal point the 'u', 'n' or 'p' is sometimes put in place of the decimal point. Therefore a  $4.7\text{pF}$  capacitor can be marked as 4p7. If no unit is given, a judgement, based on the capacitor's physical size, must be made to determine which unit is intended. For example, a small ceramic capacitor marked '4.7' is probably  $4.7\text{pF}$ , whereas a large plastic capacitor marked '4.7' is more likely to be  $4.7\mu\text{F}$ . If the value is in nF then this is invariably shown.

Another marking system uses 3 numeric digits to indicate the value in picofarads. The first two digits represent the first two digits of the value and the third digit is the multiplier or number of zeroes.

For example, a capacitor marked 104 would be read as 1, 0, 0000. This would be formatted as  $100,000\text{pF}$  and would commonly be known as  $100\text{nF}$  or  $0.1\mu\text{F}$ . Likewise a capacitor marked 472 would be  $4700\text{pF}$ , also known as  $4.7\text{nF}$  or  $.0047\mu\text{F}$ .

A similar system represents these 3 digits using colours taken from the resistor code, instead of numbers.

### SOME COMMON VALUES AND THEIR POSSIBLE MARKINGS:

MICROFARADS	NANOFARADS	PICOFARADS	EIA CODE
0.0001 $\mu\text{F}$ *	0.1n*	100pF	101
0.00022 $\mu\text{F}$ *	0.22n (n22)	220pF	221
0.001 $\mu\text{F}$	1n (1n0)	1,000pF	102
0.0033 $\mu\text{F}$	3.3n (3n3)	3,300pF	332
0.01 $\mu\text{F}$	10n	10,000pF*	103
0.047 $\mu\text{F}$	47n	47,000pF*	473
0.1 $\mu\text{F}$ (u1)	100n	100,000pF*	104
0.82 $\mu\text{F}$ (u82)	820n	820,000pF*	824
1.0 $\mu\text{F}$ (u0)	1000n*	1,000,000pF*	105

\* Not normally expressed in this form.

### VOLTAGE RATING

Voltage rating is usually marked and is often identified by the symbol 'V'. Most electrolytic capacitors clearly indicate their voltage rating. Polyester capacitors usually show the voltage rating but often omit the 'V' symbol. Small ceramic capacitors often show no voltage rating.

If the capacitance and voltage rating are both marked, a unit is also marked for at least one of the quantities so that the two cannot be confused.

### TOLERANCE

Tolerance indicates how close a capacitor's actual value is likely to be to its marked value.

A tolerance can be marked numerically, as a code consisting of a single letter, or, on colour-coded capacitors, as a fourth coloured band. The code letter is usually placed immediately after the value. Commonly used tolerance codes are:

CODE	COLOUR	TOL	CODE	COLOUR	TOL
A	-	+20-10 (2)	L	-	+/-15%
C	red	+/-0.25pF (1)	M	black	+/-20%
D	green	+/-0.5pF (1)	N	-	+/-30%
E	white	+/-1.0pF (1)	P	-	+100-0
F	brown	+/-1%	Q (2)	-	+30 -10
G	red	+/-2%	S	-	+50 -20
J	green	+/-5%	W	-	+50 -10
K	white	+/-10%	W (2)	-	+40 -20
	Z		grey	+80 -20	

(1) used on capacitors  $\leq 10\text{pF}$   
(2) used on electrolytic capacitors

### POLARITY

Polarity sensitive capacitors, such as electrolytics, are usually marked with a '+' or '-' symbol adjacent to one lead to indicate polarity. Thomson brand tantalum capacitors may have a triangular logo to indicate the positive lead, instead of the '+' symbol.

## TEMPERATURE CHARACTERISTICS

All real capacitors exhibit some change in value with varying temperature. Some ceramic types exhibit fairly linear changes and are useful as temperature compensating elements in AC circuits. The temperature coefficients of these types may be marked in letter codes or designated by a coloured spot.

TEMPCO CODE	EIA CODE	JIS CODE	TEMPCO COLOUR	PPM/°C
P100			red/violet	+100
NP0	COG	C	black	0
N30	S1G	H		-30
N033	S1G		brown	-33
N075	U1G		red	-75
N080	U1G	L	red	-80
N150	P2G	P	orange	-150
N220	R2G	R	yellow	-220
N330	S2H	S	green	-330
N470	T2H	T	blue	-470
N750	U2J	U	violet	-750
N1500	P3K	W	orange/orange	-1500
N2200	R3L			-2200
P350/N1000	SL	SL		+350 to -1000

Capacitors using the JIS code sometimes have a second letter to designate the temperature coefficient's tolerance.

LETTER	TEMPCO TOLERANCE
G	+/-30ppm/°C
H	+/-60ppm/°C
J	+/-120ppm/°C
K	+/-250ppm/°C
L	+/-500ppm/°C

For example, a capacitor marked 'CH' would have a temperature coefficient of between  $+60\text{ppm/°C}$  and  $-60\text{ppm/°C}$  ('C'=0, 'H' = +/-60ppm/°C).

Ceramic capacitors with non-linear temperature coefficients sometimes use a 3-digit code to indicate – their operating temperature range and their stability over that range.

The 1st character indicates minimum operating temperature, the 2nd, maximum temperature and the 3rd gives the stability over this temperature range.

1ST CHARACTER	MIN TEMP °C	2ND CHARACTER	°C	MAX
X	-55	5		+85
Y	-30	7		+125
Z	+10			

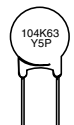
3RD CHARACTER	STABILITY (%)
F	+/-7.5
P	+/-10
R	+/-15
S	+/-22
T	+22/-33
U	+22/-56
V	+22/-82

An example of this is the common 'Z5U' type used in bypass applications. This capacitor operates over the  $+10^\circ$  to  $+85^\circ$  temperature range and exhibits a stability of  $+22$  to  $-56\%$  over this temperature range.

### PUTTING IT ALL TOGETHER

Knowing how the important information is likely to be marked, we can decode the markings on a capacitor and determine its value, voltage rating, tolerance and sometimes its temperature characteristic.

For example, a capacitor marked 104K 63V Y5P will be  $0.1\mu\text{F}$  (decoded from the 104) having a +/-10% tolerance (decoded from the K), a 63 Volt rating, an operating temperature range of  $-30^\circ$  to  $+85^\circ$  (decoded from Y5) and a stability of +/-10% (P) over this range. Likewise, 6n8K63 would indicate  $6.8\text{nF}$  (from 6n8), +/-10%, (from K) and 63 Volts (from 63).



104 = 1 0 0000 pF  
 = 100nF = 0.1  $\mu\text{F}$   
 K =  $\pm 10\%$  Tolerance  
 63 = 63V  
 Y =  $-30^\circ\text{C}$   
 5 =  $+85^\circ\text{C}$   
 P =  $\pm 10\%$  Stability

### FILM CAPACITOR TYPES

KP	Polypropylene film/foil
KS	Polystyrene film/foil
KT	Polyester film/foil
	(polyethylene terephthalate PETP)
MK	Metallised plastic film (general)
MKC	Metallised polycarbonate
MKP	Metallised polypropylene
MKT	Metallised polyester
	(polyethylene terephthalate, PETP)
MKT-P	Metallised polyester/paper
MKY	Metallised low-loss polypropylene
MKL (MKU)	Metallised lacquer (cellulose acetate)

## CAPACITORS AND THEIR USES

### ALUMINIUM ELECTROLYTIC CAPACITORS

Ideal for use in filtering or smoothing applications in power supplies. Also used for coupling and bypassing in audio circuits and as a timing element in non critical circuits.

Modern aluminium electrolytic capacitors have high reliability and low leakage. Special low leakage versions are available with leakage which rivals that of tantalum capacitors.



Single ended (- end) (+ end)

### SOLID TANTALUM CAPACITORS

Offer smaller size and lower leakage than standard aluminium electrolytics.

### CERAMIC CAPACITORS

Offer low cost and high capacitance in a small physical volume. There are generally two types:

High stability, temperature compensating types for use in resonant circuit and filter applications. These have linear temperature characteristics, and their value is largely independent of voltage and frequency.

Bypass and coupling capacitors for use in less critical applications. These are less stable, have non-linear temperature characteristics and are somewhat voltage dependent.

### PAPER CAPACITORS

An original construction, now rarely used. Plastic film capacitors have replaced paper capacitors in most applications.

### POLYESTER CAPACITORS

Low cost, good stability and available in a large range of values. These are the most widely used capacitors for general purpose applications.

Greencaps and MKT type capacitors are examples of polyester (polyethylene terephthalate) film capacitors.

### POLYCARBONATE CAPACITORS

Offer low temperature coefficient and lower dielectric losses at high frequency. Most often chosen for temperature stability characteristics.

### POLYSTYRENE CAPACITORS

Usually chosen for applications requiring tight tolerance coupled with high stability. Predictable temperature coefficient used in conjunction with particular ferrite cores makes highly stable tuned circuits or oscillators.

### POLYPROPYLENE CAPACITORS

Offer very low dielectric losses and good temperature coefficient. Used in power electronics applications, e.g. mains capacitors, switching power supplies, inverters and TV deflection circuits.

### INDUCTORS

There are a variety of marking systems for small inductors. The value may be marked directly, with a three digit numeric code or a colour code. The tolerance may be marked directly, with a single letter, or a colour code.

The 3-digit numeric code indicates the value in microhenries. The first two digits represent the first two digits of the value and the third is the multiplier, or number of zeroes. For example, the code 472 would be interpreted as  $47\text{00}\mu\text{H}$ , or  $4.7\text{mH}$ . Likewise, 103 would represent  $10\text{mH}$ .

Where coloured bands are used, the system is essentially the same as that used for the resistor colour code, but with the value expressed in microhenries. An exception is that some inductors include an extra, thicker band at one end to identify milspec characteristics.

Tolerance codes are usually placed after the value code. M is used to represent +/-20%, K or a silver band for +/-10%, J or a gold band for +/-5%.

Resin-dipped radial lead types are often marked with a sequence of coloured dots. The significance of these dots is shown below.



4th Dot (Tolerance)  
 1st Dot  
 2nd Dot  
 3rd Dot (Multiplier)

## CAPACITOR CODE AND CONVERSION CHART

MICROFARAD	NANOFARAD.	PICOFARAD	EUROPEAN	EIA/MIL (USA)	COLOUR BAND CODE
0.001uF	1.0nF	1000pF	1n0	102	Brown-Black-Red
0.0012uF	1.2nF	1200pF	1n2	122	Brown-Red-Red
0.0015uF	1.5nF	1500pF	1n5	152	Brown-Green-Red
0.0018uF	1.8nF	1800pF	1n8	182	Brown-Grey-Red
0.0022uF	2.2nF	2200pF	2n2	222	Red-Red-Red
0.0027uF	2.7nF	2700pF	2n7	272	Red-Violet-Red
0.0033uF	3.3nF	3300pF	3n3	332	Orange-Orange-Red
0.0039uF	3.9nF	3900pF	3n9	392	Orange-White-Red
0.0047uF	4.7nF	4700pF	4n7	472	Yellow-Violet-Red
0.0056uF	5.6nF	5600pF	5n6	562	Green-Blue-Red
0.0068uF	6.8nF	6800pF	6n8	682	Blue-Grey-Red
0.0082uF	8.2nF	8200pF	8n2	822	Grey-Red-Red
0.01uF	10nF	10x10 <sup>3</sup> pF*	10n	103	Brown-Black-Orange
0.012uF	12nF	12x10 <sup>3</sup> pF*	12n	123	Brown-Red-Orange
0.015uF	15nF	15x10 <sup>3</sup> pF*	15n	153	Brown-Green-Orange
0.018uF	18nF	18x10 <sup>3</sup> pF*	18n	183	Brown-Grey-Orange
0.022uF	22nF	22x10 <sup>3</sup> pF*	22n	223	Red-Red-Orange
0.027uF	27nF	27x10 <sup>3</sup> pF*	27n	273	Red-Violet-Orange
0.033uF	33nF	33x10 <sup>3</sup> pF*	33n	333	Orange-Orange-Orange
0.039uF	39nF	39x10 <sup>3</sup> pF*	39n	393	Orange-White-Orange
0.047uF	47nF	47x10 <sup>3</sup> pF*	47n	473	Yellow-Violet-Orange
0.056uF	56nF	56x10 <sup>3</sup> pF*	56n	563	Green-Blue-Orange
0.068uF	68nF	68x10 <sup>3</sup> pF*	68n	683	Blue-Grey-Orange
0.082uF	82nF	82x10 <sup>3</sup> pF*	82n	823	Grey-Red-Orange
0.1uF	100nF	10x10 <sup>4</sup> pF*	μ10	104	Brown-Black-Yellow
0.12uF	120nF	12x10 <sup>4</sup> pF*	μ12	124	Brown-Red-Yellow
0.15uF	150nF	15x10 <sup>4</sup> pF*	μ15	154	Brown-Green-Yellow
0.18uF	180nF	18x10 <sup>4</sup> pF*	μ18	184	Brown-Grey-Yellow
0.22uF	220nF	22x10 <sup>4</sup> pF*	μ22	224	Red-Red-Yellow
0.27uF	270nF	27x10 <sup>4</sup> pF*	μ27	274	Red-Violet-Yellow
0.33uF	330nF	33x10 <sup>4</sup> pF*	μ33	334	Orange-Orange-Yellow
0.39uF	390nF	39x10 <sup>4</sup> pF*	μ39	394	Orange-White-Yellow
0.47uF	470nF	47x10 <sup>4</sup> pF*	μ47	474	Yellow-Violet-Yellow
0.56uF	560nF	56x10 <sup>4</sup> pF*	μ56	564	Green-Blue-Yellow
0.68uF	680nF	68x10 <sup>4</sup> pF*	μ68	684	Blue-Grey-Yellow
0.82uF	820nF	82x10 <sup>4</sup> pF*	μ82	824	Grey-Red-Yellow

### Notes:

\*These values are not normally expressed in nanofarads (nF) or picofarads (pF). Electrolytics are usually marked in microfarads (uF) Capacitor colour code follows a similar system to that used for resistors and is expressed in picofarads. Colour Coding may vary depending on the type of capacitor, age and manufacturers preference. Most modern capacitors are marked using the European or EIA codes.

## CAPACITOR CODE AND CONVERSION CHART

MICROFARAD	NANOFARAD.	PICOFARAD	EUROPEAN	EIA/MIL (USA)	COLOUR BAND CODE
1.0uF	1000nF	10x10 <sup>5</sup> pF*	1μ0	105	Brown-Black-Green
1.2uF	1200nF	12x10 <sup>5</sup> pF*	1μ2	125	Brown-Red-Green
1.5uF	1500nF	15x10 <sup>5</sup> pF*	1μ5	155	Brown-Green-Green
1.8uF	1800nF	18x10 <sup>5</sup> pF*	1μ8	185	Brown-Grey-Green
2.2uF	2200nF	22x10 <sup>5</sup> pF*	2μ2	225	Red-Black-Green
2.7uF	2700nF	27x10 <sup>5</sup> pF*	2μ7	275	Red-Red-Green
3.3uF	3300nF	33x10 <sup>5</sup> pF*	3μ3	335	Orange-Orange-Green
3.9uF	3900nF	39x10 <sup>5</sup> pF*	3μ9	395	Orange-White-Green
4.7uF	4700nF	47x10 <sup>5</sup> pF*	4μ7	475	Yellow-Violet-Green
5.6uF	5600nF	56x10 <sup>5</sup> pF*	5μ6	565	Green-Blue-Green
6.8uF	6800nF	68x10 <sup>5</sup> pF*	6μ8	685	Blue-Grey-Green
8.2uF	8200nF	82x10 <sup>5</sup> pF*	8μ2	825	Grey-Red-Green
10.0uF	10000nF	10x10 <sup>6</sup> pF*	10μ	106	Brown-Black-Blue
15.0uF	15000nF	15x10 <sup>6</sup> pF*	15μ	156	Brown-Green-Blue
18.0uF	18000nF	18x10 <sup>6</sup> pF*	18μ	186	Brown-Grey-Blue
22.0uF	22000nF	22x10 <sup>6</sup> pF*	22μ	226	Brown-Black-Blue
27.0uF	27000nF	27x10 <sup>6</sup> pF*	27μ	276	Red-Red-Blue
33.0uF	33000nF	33x10 <sup>6</sup> pF*	33μ	336	Orange-Orange-Blue
39.0uF	39000nF	39x10 <sup>6</sup> pF*	39μ	396	Orange-White-Blue
47.0uF	47000nF	47x10 <sup>6</sup> pF*	47μ	476	Yellow-Violet-Blue
56.0uF	56000nF	56x10 <sup>6</sup> pF*	56μ	566	Green-Blue-Blue
68.0uF	68000nF	68x10 <sup>6</sup> pF*	68μ	686	Blue-Grey-Blue
82.0uF	82000nF	82x10 <sup>6</sup> pF*	82μ	826	Grey-Red-Blue
100uF	100,000nF	10x10 <sup>7</sup> pF*	100μ	107	Brown-Black-Violet
220uF	220,000nF	22x10 <sup>7</sup> pF*	220μ	227	Red-Red-Violet

### CAPACITOR TOLERANCES

Brown	= +/-1%
Red	= +/-2%
Green	= +/-5%
White	= +/-9% or +/-10%
Black	= +/-20%
F	= +/-1%
G	= +/-2%
J	= +/-5%
K	= +/-10%
M	= +/-20%

### CAPACITOR VOLTAGES

(Flat Film Type, Last Band)	
Brown	= 100 VDC
Red	= 250 VDC
Yellow	= 400 VDC
Violet	= 630 VDC
(Solid Tantalum, Last Band)	
Black	= 10 VDC
Brown	= 1.6 VDC
Red	= 4.0 VDC
Orange	= 40 VDC
Yellow	= 400 VDC
Green	= 16 VDC
Silver	= 25 VDC
White	= 2.5 VDC
Pink	= 35 VDC

### CAPACITANCE FORMULAE

Parallel:  $CT = C1 + C2 + C3 +$

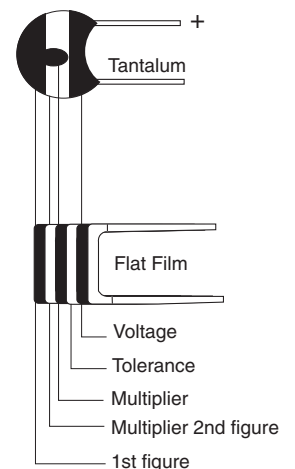
$$\text{Series: } CT = \frac{C1 \times C2}{C1 + C2}$$

$$CT = \frac{1}{\frac{1}{C1} + \frac{1}{C2} + \frac{1}{C3} + \dots}$$

Stored Charge:  $Q = C \times E$

$$\text{Stored Energy: } W = \frac{C \times E^2}{2}$$

Q = Charge in Coulombs  
 C = Capacitance in Farads (F)  
 E = Voltage (V)  
 W = Energy in Joules (watt-seconds)



WIRE AND COAXIAL CABLE SPECIFICATIONS

Table with columns: Type, Nom. Imp Zo, Outer Dia. (mm), Nom. Attenuation (dB/100m) at 50, 100, 200, 400 MHz and 1 GHz, Vel. Factor %, Cap. pF/m, Volt. Vrms.

Note: Except for Belden 9913, all specifications are referenced to MIL-C-170/D/F or JAN-C-17A as applicable.

WINDING WIRE DATA

Table with columns: B&S Gauges, SWG, Metric mm, Dia. mils 1000ft, Ohms per, Dia. mm km, Ohms per.

NOTE: Dick Smith Electronics does not stock all cables and wires shown. Data is presented for comparative purposes.

COPPER CABLE CHARACTERISTICS

Table with columns: Number & Size of Strands, Nominal Conductor Area (sq. mm), Nominal Current Rating (Amps), Maximum Resistance per metre (ohms 35°C), Nearest Equivalent AWG (B&S).

Note: The nominal current ratings are intended as guidelines for low power, electronics, communications and control applications only.

PROPERTIES OF COPPER

(Based upon the IEC International Annealed Copper Standard.)

- Volume resistivity @ 20°C: 0.0000017241 ohm-cm
Mass resistivity @ 20°C: 0.15328 ohm-grams/sq. metre
Temperature coefficient @ 20°C: 0.0039
Tensile strength: 2350 kg/sq. cm
Melting point: 1083°C



CHOOSE THE RIGHT CABLE

The following 6 steps should help when trying to choose a suitable wire size for hooking up marine and auto accessories, extension speakers and the like.

- 1 Decide how much of the available supply voltage you are prepared to lose. All practical wires have some finite resistance, so it is impossible not to lose some voltage.
2 Divide the figure obtained in (1) by the maximum current which you expect to draw through the wire. This result is the tolerable resistance of your wiring.
3 Divide the result obtained in (2) by the total length (in metres) of the wiring loop. Include the distance from the power source to the load, and back again. This result is the maximum resistance of the desired wire in ohms per metre.
4 From the chart above, select a wire which has a 'resistance per metre' figure which is equal to or less than the desired value.
5 Check that the 'nominal current rating' in the table is not exceeded. If it is, then you may need to select a larger wire size with an adequate rating. While the figures given are conservative, they must not be exceeded in 240V applications and may need derating in higher temperature situations.

6 Finally, when purchasing the wire, verify that its insulation is rated to withstand the voltage and temperature for your particular application.

EXAMPLE: You wish to install a 12V, 55W light on a boat. The current drawn by the light will be (55/12) 4.6 amps. You calculate that the length of the wiring loop is 5 metres.

- Step 1 Lets say that you'll tolerate a loss of 5% of the available 12V supply, i.e. 0.6V.
Step 2 Divide 0.6V by 4.6A to obtain 0.13 ohms.
Step 3 Divide 0.13 ohms by 5 metres to obtain 0.026 ohms per metre.
Step 4 Referring to the chart, you will find that 24 x 0.2mm wire appears to be suitable as its resistance is 0.025 ohms per metre.
Step 5 The 'nominal current rating' for this cable is 7.5A and on this basis it is still suitable.
Step 6 As you are dealing with a 12V system and are able to keep the wire away from hot areas and protect it from abrasion, there is unlikely to be any problem related to the insulation.

BIPOLAR TRANSISTORS

Note: Some small signal transistors may have a TO-92 case and a 'PW' prefix. The electrical specifications are the same, only the case is changed. See case outline TO-92(72) for pin detail. Not all devices in this section are stocked by Dick Smith Electronics.

Table with columns: TYPE MAT, CASE, POL, VCE mA, Vcb, Ic, VCES @ Ic, hFe @ Ic, MHz, Ptot mW, USE, COMPARABLE TYPES. Lists various transistor models and their specifications.

TRIACS

Table with columns: Type, CASE, Vdrm, It (rms) (A), Ifsm (A), Igt (mA), Vgt (T = 25°C), Pg (av) (W), Ih (mA), Comments (25°C), dv/dt (V/us). Lists triac models and their specifications.

## FETS

Type	Case	BVgss V @	Ig (uA)	Vgs (off) Min Max @ Vds	Id (nA)	Idss (mA) Min Max @ Vds	Vgs	YFS (umhos) Min Max @ Vds	Ptot mW	Use/Comments	
BF245B	TO-92 Var 1	30	1	0.5 8	15 10	6 15	15 0	3000	6500	15 300	N/CH Junction audio to H.F.
BF981	SOT-103	>6		2.5 10	20mA	4 25	10 0		14000 typ.	225	N/CH Dual gate MOS. VHF AMP.
BF998	SMT	>6	±10	2.5 8	20mA	2 18	10 0/4		24000 typ.	8 200	N/CH VHF/UHF dual gate SMT
BFR84	TO-72(2)	>6	100	1.5 3.8	10	20 55	10 0		15000 typ.	10 300	N/CH VHF dual gate
MFE131	TO-72(2)	±6	±10	4 15	200mA	30 15	0	8000	20000	15 300	N/CH Dual gate MOS. VHF amp.
MPF102	TO-92(72)	25	10	0.5 8	15 2	2 20	15 0	2000	7500	15 310	N/CH Junction – VHF
MPF105	TO-92(72)	25	1	8 15	10	4 16	15 0	2000	6000	15 310	N/CH Junction – audio Sw.
MPF106	TO-92(72)	25	1	0.5 4	15 10	4 10	15 0	2500	7000	15 310	N/CH Junction – RF
MPF131	262	±6	±10	4 15	200mA	3 30	15 0	8000	20000	15 350	N/CH Dual gate MOS. VHF amp
2N4342	TO-92(72)	25	10	5 10	10	12 30	10 0		6000	10 180	P/CH Junction – audio Sw.
2N5245	TO-92(77)	30	1	1.5 6	10	1 18	0		5800 typ.	10 360	N/CH VHF/UHF MIXER
2N5247	to-92(72)	25	10	0.5 6	15 10	1 5	15 0	1000	5000	15 625	n/ch Junction – audio Sw.
2N5459	TO-92(72)	25	1	2 8	15 10	4 9	15 0	2000	6000	15 310	N/CH Junction – audio Sw.
2N5484	TO-92(72)	25	1	0.3 3	15 10	1 5	15 0	3000	6000	15 310	N/CH Junction – VHF
2N5485	TO-92(72)	25	1	0.5 4	15 10	4 10	15 0	3500	7000	15 310	N/CH Junction – VHF
2N5486	TO-92(72)	25	1	2 6	15 10	8 20	15 0	4000	8000	15 310	N/CH Junction – VHF
3SK40	TO-72(2)	±6	±10	4 15	200mA	25 15	0	8000		15 250	N/CH Dual gate MOS. VHF amp.
3SK121	2-6F1A	-5		-2.5 4	5 100mA	20 45	5 0		17000 typ.	10 200	N/CH UHF GaAs dual gate

## POWER FETS

Type	Ptot Case	Vds (max) (V)	Id (max) (V)	Vgs (thres.) (A)	Gfs min	Ciss max	Rds umhos	(pF)	(ohms)	Comment
BUK456-60A	TO-220	150	60	52	2.1	4	17M	2000	0.028	Inverters
BUK456-60A	TO-220	150	60	52	2.1	4	17M	2000	0.028	Inverters
BUK457-600B	TO-220	150	600	7.1	2.1	4	8M typ.	1800	1.2	Mains SMPS
BUZ71A	TO-220	40	50	13	2.1	4	4M	600	0.12	Nch power MOSFET
IRF520	TO-220	60	100	9.2	2	4	4M	350	0.25	Nch power MOSFET
IRFp450	to-247	180	500	14	2.0	4.0	14m	2700	0.4	Nch power MOSFET
MTP3055E	TO-220	40	60	12	2	4.5	4M	500	0.15	Nch power MOSFET
php6n60e	to-220	125	600	6.5	3.0	4.0	5m	1500	1.2	Nch power MOSFET
VN10K	TO-92(76)	1	60	0.3	0.3	2.5	100K	48	5	Nch DMOS FET
VN88AF	TO-202F	12.5	80	2	0.8	2				Nch VMOS FET
2SJ48	TO-3	100	120	7	0.15	1.45	1M	900	1	Pch power MOSFET
2SJ49	TO-3	100	140	7	0.15	1.45	1M	900	1	Pch power MOSFET
2SJ162	TOP-3	100	160	7	0.15	1.45	1M	900	1	Pch power MOSFET
2SK133	TO-3	100	120	7	0.15	1.45	1M	600	1	Nch power MOSFET
2SK134	TO-3	100	140	7	0.15	1.45	1M	600	1	Nch power MOSFET
2SK1058	TOP-3	100	160	7	0.15	1.45	1M	600	1	Nch power MOSFET
60n06	to-220	150	60	60	2.0	4.0	20M	1950	0.014	Nch power MOSFET

## SMALL SIGNAL DIODES

Type	Case	Vr	If (ma)	Cd (pF)	Vf	@	If (mA)	Ir (ua)	@	Vr	Trr (ns)	Use / Comparable Types
<b>Germanium</b>												
OA47	DO-7	25	110	3.5	0.45		10	100		25	70	Gold bonded G.P. switching
OA90	DO-7	20	45		1.5		10	450		20		G.P. point contact OA70, OA80
OA91	DO-7	90	50		1.9		10	180		75		G.P. point contact OA71, 79, 81
DSOA91	DO-7	50	30		1		5	200		10		G.P. OA91, 1N60
1N60	DO-7	40	30		0.05		0.375	200		10		AM/FM detector
OA95	DO-7	90	50		1.5		10	110		0.75		G.P. point contact
<b>Silicon</b>												
BA102	DO-7	20		20-45	Cd Ratio 1.4	@	4V/10V					Variable capacitance
BA234/4	DO-35	20	100	<2	1		100	0.1		15		UHF Sw.
BAW62	SOD-27	75	200		2	@	5	5		75	4	High speed silicon Sw.
BB119	DO-35	15		20-25 @ 4V	Cd Ratio >1.3	@	4V/10V					Varicap – replaces BA102
BB122	DO-35	30		12 @ 3V	Cd Ratio 5.2	@		0.05		28		Varicap VHF UHF
BB212	TO-92	12		500-620 @ 0.5V	Cd Ratio >2.5	@	0.5V/8V	0.05		10		AM dual varicap
1N914A	DO-35	75	75	4	1		10	5		75	4	Small signal Sw. 1N4148
1N4148	DO-35	75	200	4	1		10	0.025		20	4	Small signal Sw. 1N914A
1N4448	DO-35	75	300	4	1		100	0.025		6	4	G.P. silicon Sw. 1N4148
5082-2800	DO-7	70	15	2	0.41		1	0.2		50	0.1	Schottky, UHF detector, mixer switch

## RECTIFIERS

Type	Vr	If (a)	Ifsm (a)	Vf	@	If (A)	Ir (ua)	@	Vr	Use
BYX98/600	600	10	75	1.4		10	200		600	Power rectifier stud mount
1N4004	400	1	30	1.1		1	5		400	G.P. rectifier
1N4007	1000	1	30	1.1		1	5		1000	G.P. rectifier
BY229-400	400	7	60	1.85		20				SMPS trr = 150nS
1N4936	400	1	30	1.3		1	5		400	SMPS trr = 200nS
1N5404	400	3	200	1.1		3	5		400	G.P. rectifier
1N5408	1000	3	200	1.1		3	5		1000	G.P. rectifier
P600M	1000	6	400	0.9		6	25		1000	G.P. rectifier
6a10	1000	6	350	1.0		6	10		1000	G.P. rectifier
FB5006	600	50	500	1.2 (per leg)		25	10		600	High current bridge
3504	400	35	400	1.2 (per leg)		17.5	10		400	High current bridge
PD4	400	6		1.3 (per leg)		6	10		400	Medium duty mini bridge
KBL04	400	4	200	1.1 (per leg)		3	10		400	G.P. bridge
KBPC804	400	8	200	1.0 (per leg)		4	10		400	High current bridge
KBL10/407	1000	4	200	1.1 (per leg)		2	10		1000	G.P. bridge SIL pins
W04	400	1.5	50	1.1 (per leg)		1	10		400	Mini bridge
W06	600	1.5	30	1.1 (per leg)		1	10		600	G.P. bridge
2W06	600	2	50	1.0 (per leg)		1	10		600	G.P. bridge
DB155G	600	1.5	50	1.1 (per leg)		1	10		600	Mini bridge DIL pins
1N5819	40	1	25	0.6		1	1mA		40	Schottky barrier
1N5822	40	3	80	0.525		3	2mA		40	Schottky barrier

## SCRs

Type	Case	VDRM	IT (RMS) (A)	IFSM (A)	IGT (mA)	VGT	Pc (AV) (w)	dv/dt (V/uS)	COMMENTS (25°C)
BT151-500R	TO-220	500	12	100	15 max	1.5 max	0.5	20	max 200 (RGK=100)
BT169B	TO-92A	200	0.8	8	2 max	0.8 max	0.1	5	max 25 (RGK=1k)
C103B	TO-92	200	0.8	8	2 max	0.8 max	0.01	5	max 20 (RGK=1k)
C106Y	TO-202	30	4	20	2 max	0.8 max	0.1	5	max 8 (RGK=1k)
C106D	TO-126/TO-202	400	4	20	2 max	0.8 max	0.1	3	max 8 (RGK=1k)
C122D	TO-220	400	8	82	25 max	1.5 max	0.5	3	max 50 (Gate o/c)
C122E	TO-220	500	8	82	25 max	1.5 max	0.5	30	max 50 typ (Gate o/c)
C203B	TO-92A	200	0.8	8	2 max	0.8 max	0.01	5	max 20 typ
C220D	TO-48	400	10	82	25 max	1.5 max	0.5	30	max 50 typ
P0103AA	TO-92A	100	0.8		0.2-2	0.8 max	0.1	5	max 100 typ
P0103AB	TO-92	100	0.8		0.2-2	0.8 max	0.1	5	max 100 typ
S4008	TO-220	400	8	85	25 max	1.5 max	0.5	30	max 50 typ (Gate o/c)
S6008L	TO-220	600	8	85	25 max	1.5 max	0.5	30	max 50 typ (Gate o/c)
SPS420	TO-48	400	20	200	25 max	2.0 max	0.5	50	max 200 (Gate o/c)
TYN608	TO-220	600	8	80	15 max	1.5 max	-	30	max 200 typ
TYN608G	TO-220	600	8	80	25 max	1.5 max	-	45	max 500
tyr816	to-220	800	16	160	25 max	1.5 max	1	40	max 500 min (Gate o/c)



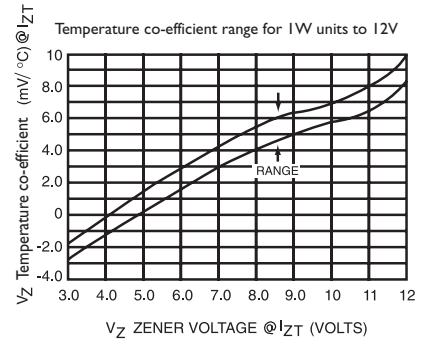
## ZENER DATA

Type No. ▲ 400mW Series	Nominal Type No. ▼ 1 Watt Series	Test Current Zener Volt. V <sub>Z</sub> @I <sub>ZT</sub> (V)	I <sub>ZT</sub> (mA) 400mW	1W
1N746	1N4728	3.3	20	
1N747	1N4729	3.6	20	
1N748	1N4730	3.9	20	
1N750	1N4732	4.7	20	53
1N751	1N4733	5.1	20	49
1N752	1N4734	5.6	20	45
1N753	1N4735	6.2	20	41
1N754,1N957	1N4736	6.8	18.5	37
1N755,1N958	1N4737	7.5	16.5	34
1N756,1N959	1N4738	8.2	15	31
1N757,1N960	1N4739	9.1	14	28
1N758,1N961	1N4740	10	14	25

TYPE NO. ▲ 400MW SERIES	NOMINAL TYPE NO. ▼ 1 WATT SERIES	TEST CURRENT ZENER VOLT. V <sub>Z</sub> @I <sub>ZT</sub> (V)	I <sub>ZT</sub> (MA) 400MW	1W
1N962	1N4741	11	12.5	23
1N759, 1N963	1N4742	12	11.5	21
1N964	1N4743	13	10.5	19
1N965	1N4744	15	9.5	17
1N966	1N4745	16	8.5	15.5
1N967, 1N4112	1N4746	18	7.8	14
1N968	1N4747	20	7	12.5
1N969	1N4748	22	6.2	11.5
1N970	1N4749	24	5.6	10.5
1N971	1N4750	27	5.3	9.5
1N972	1N4751	30	5.3	8.5
1N973	1N4752	33	5.2	7.5

▲ (DC Power Dissipation: 400 milliwatts @ 50°C Ambient)  
(Derate 3.2mW/°C above 50°C)

▼ (DC Power Dissipation: 1W @ 50°C Ambient) (Derate 6.67mW/°C above 50°C). This range will dissipate up to 3W @ 75°C with 10mm lead length as heatsink.



## DISCRETE OPTOELECTRONICS

### 7-SEGMENT LEDs

Z 4118    Z 4104    Z 4146    Z 4118    Z-4130    Z4146

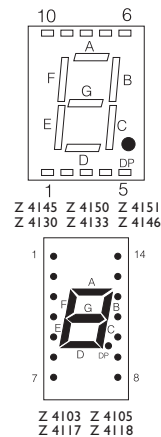
#### Characteristics

	Z 4118	Z 4104	Z 4146	Z 4118	Z-4130	Z4146
Common terminal pol.	Anode	Cathode	Anode	Anode	Cathode	Anode
Digit size	7.6mm	7.6mm	13.2mm	7.6mm	14.2mm	13.1mm
Colour	Red	Red	Red	Orange	Red	Red
Av. fwd. seg. current	25mA	25mA	25mA	25mA	25mA	25mA
Segment voltage	1.7	1.7	1.7	2.0	1.7	
Min. rev. brkdn. volt.	5	5	5	5	5	
Max. rev. current	100uA	100uA	100uA	10uA	10uA	
Seg. intensity (typ.)	450ucd	450ucd	500ucd	750ucd	4500ucd	8500ucd
Max. seg. dissipation	55mW	55mW	55mW	85mW	100mW	110mW

#### Connections

Seg.	Z 4118	Z 4104	Z 4146	Z 4118	Z-4130	Z4146
Seg. A	1	10	7	1	7	7
Seg. B	13	9	6	13	6	6
Seg. C	10	8	4	10	4	4
Seg. D	8	5	2	8	2	2
Seg. E	7	4	1	7	1	1
Seg. F	2	2	9	2	9	9
Seg. G	11	3	10	11	10	10
Dec. Pt.	9	7	5	9	5	5
Common	3, 14	1, 6	3, 8	3, 14	3, 8	3, 8

Note: Not all devices in this section are stocked by Dick Smith Electronics.

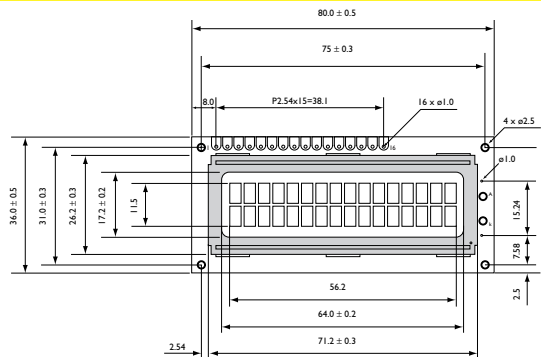


## Z 4170 / Z 4172 LIQUID CRYSTAL DISPLAYS

The Z 4170 and Z 4172 are self-contained Liquid Crystal Displays (LCDs). Both modules can display 2 rows of 16 characters, and both include a CMOS interface and drive IC to enable simple connection to a 4 or 8-bit microprocessor. 96 alpha-numeric characters are available from an in-built character generator, and up to 8 characters can be user defined. The Z 4172 includes a backlight feature. A full data sheet is included with each unit.

#### Interface Pin Functions

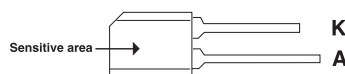
Pin No	Symbol	Function
1	Vss	Ground (0V)
2	Vdd	Logic supply (+5V)
3	VO	Contrast adjustment
4	RS	Data/Instruction select
5	R/W	Read/write select
6	E	Signal enable
7-14	D0-D7	Data bus
15	A	LED backlight (Z 4172 only)
16	K	LED backlight (Z 4172 only)



## Z 1956 SILICON PIN INFRARED PHOTO DIODE

Silicon PIN photo diode encapsulated in a black plastic package which acts as an infrared filter.

Reverse voltage	V <sub>r</sub> max	30V
Power dissipation @ 25°C	P <sub>tot</sub>	100mW
Junction temperature	T <sub>j</sub>	100°C
Dark reverse current (V <sub>r</sub> =10V E <sub>e</sub> =0)	I <sub>dr</sub>	30nA
Light reverse current (E <sub>e</sub> =5mW/cm <sup>2</sup> )	I <sub>lr</sub>	200uA typ
Wavelength of peak response	λ <sub>pk</sub>	940nm



## Z 3235 INFRARED EMITTING DIODE

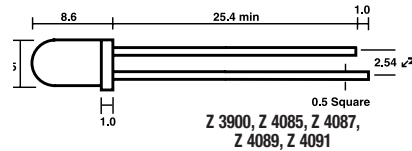
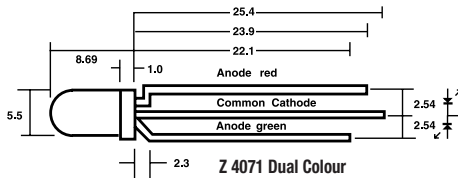
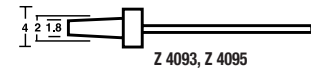
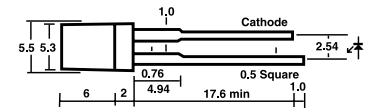
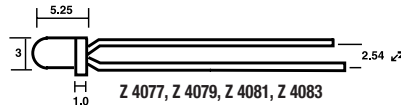
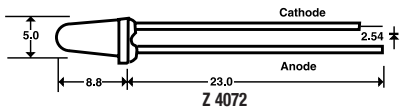
High Intensity Gallium Arsenide infrared emitting diodes intended for remote control applications. Smoke coloured end looking encapsulation.

Forward voltage	V <sub>f</sub>	1.2V @ 20mA
Continuous reverse voltage	V <sub>r</sub>	5V
Forward current (d.c.)	I <sub>f</sub> max	50mA
Peak forward current (300pps, 1us pulse)	I <sub>fpk</sub>	3A
Total power dissipation up to T <sub>amb</sub> =25°C	P <sub>tot</sub> max	100mW
Junction temperature	T <sub>j</sub>	100°C
Radiant incidence @ I <sub>f</sub> =20mA	E <sub>o</sub>	1mW/cm <sup>2</sup>
Wavelength of peak emission	λ <sub>pk</sub>	940nm



## HIGH PERFORMANCE LEDs

TYPE		NOMINAL SIZE (mm)	LENS	FWD. VOLTAGE (@ 20mA)	MAX. CONT. FWD. CURRENT (mA)	MAX. REV. VOLTAGE	LUMINOUS INTENSITY (mcd) @	CURRENT (mA)	PEAK WAVELENGTH (nm)
Z 3820	Blue	3 dia.	Water clear	3.5	30	5	700	20	468
Z 3800	White	3 dia.	Water clear	3.5	30	5	1500	20	-
Z 4077	Red	3 dia.	Red diffused	2.1	15	5	2	10	697
Z 4079	Green	3 dia.	Green diffused	2.2	30	5	18	10	565
Z 4081	Yellow	3 dia.	Yellow diffused	2.2	20	5	15	10	585
Z 4083	Orange	3 dia.	Orange diffused	2.0	30	5	15	10	635
Z 3902	Blue	5 dia.	Water clear	3.5	30	5	1800	20	475
Z 3905	Blue	5 dia.	Water clear	3.6	30	5	5600	20	475
Z 3980	White	5 dia.	Water clear	3.6	30	5	2300	20	-
Z 3982	White	5 dia.	Water clear	5	20	5	8000	30	456
Z 4015	Green	5 dia.	Water clear	3.5	30	5	14000	20	512
Z 4033	Amber yellow	5 dia.	Water clear	2.0	50	5	6500	20	595
Z 4071	Red/green	5 dia.	White diffused	2.0	25	5	126/49	20	660/567
Z 4072	Red	5 dia.	Water clear	1.8	40	4	1000	20	660
Z 4074	Red	5 dia.	Water clear	1.8	40	4	2300	20	660
Z 4031	Sunset red	5 dia.	Water clear	2.0	50	5	8000	20	640
Z 4085	Red	5 dia.	Red diffused	2.1	15	5	3	10	697
Z 4087	Green	5 dia.	Green diffused	2.2	30	5	15	10	565
Z 4089	Yellow	5 dia.	Yellow diffused	2.2	20	5	12	10	585
Z 4091	Orange	5 dia.	Orange diffused	2.0	30	5	12	10	635
Z 4044	Flashing red	5 dia.	Red	12	60	5	390	25	660
Z 4046	Flashing green	5 dia.	Green	12	60	5	98	25	567
Z 4060	Red	10 dia.	Red diffused	2.0	30	5	40	20	625
Z 4067	Red	10 dia.	Water clear	1.9	50	5	6000	20	660
Z 4200	Red	1.8 x 5.0	Red	2.1	30	5	5.1	20	690
Z 4202	Green	1.8 x 5.0	Red	2.1	30	5	27	20	565
Z 4204	Yellow	1.8 x 5.0	Yellow	2.1	30	5	27	20	565



### FLASHING LEDs

	Z 4042 GREEN	Z 4044 RED	Z 4046 GREEN	UNITS
Nominal size	5	5	5	mm
Operating voltage range	3-10	3-15	3-15	V
Intensity (typ. @ 10V)	10	390	98	mcd
Peak wavelength	565	660	567	nm
Flash rate (approx @ 10V)	2	2.4	2.4	Hz
Duty cycle	25	25	25	%
Max. rev. voltage	0.6	5	5	V

### Z 4801 LIGHT DEPENDENT RESISTOR (LDR)

RATINGS		CHARACTERISTICS	
Max. voltage	100V	Dark resistance	> 10 M ohm
Power dissipation (at 25°C)	50mW	Resistance @ 10 Lux	48-140k ohm
		Peak spectral response	560-620 nm

The approximate relationship between illumination and resistance is given by,

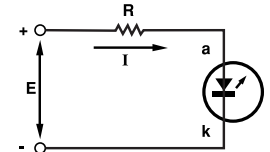
$$R = AL^{-0.85} \text{ where;}$$

R = resistance,  
A = a constant (approx. 340-990 x 10<sup>3</sup>),  
L = light level in Lux

### USING LEDs

When using a Light Emitting Diode (LED) as an indicator, use the following formula to determine series resistance for various voltages:  $R = (E - Vf) \times 1000/I$ , where R is the resistance in ohms, E is the DC supply voltage, I is the LED current in mA and Vf is the forward voltage drop of the LED, typ. 2V.

ie. With  $Vf=2V$ , LED current=20mA,  
for 6V, R=220 ohms  
9V, R=330 ohms  
12V, R=560 ohms  
24V, R=1200 ohms



Note: This does not apply to flashing LEDs. Flashing LEDs can be driven from a voltage source, ie without a resistor.

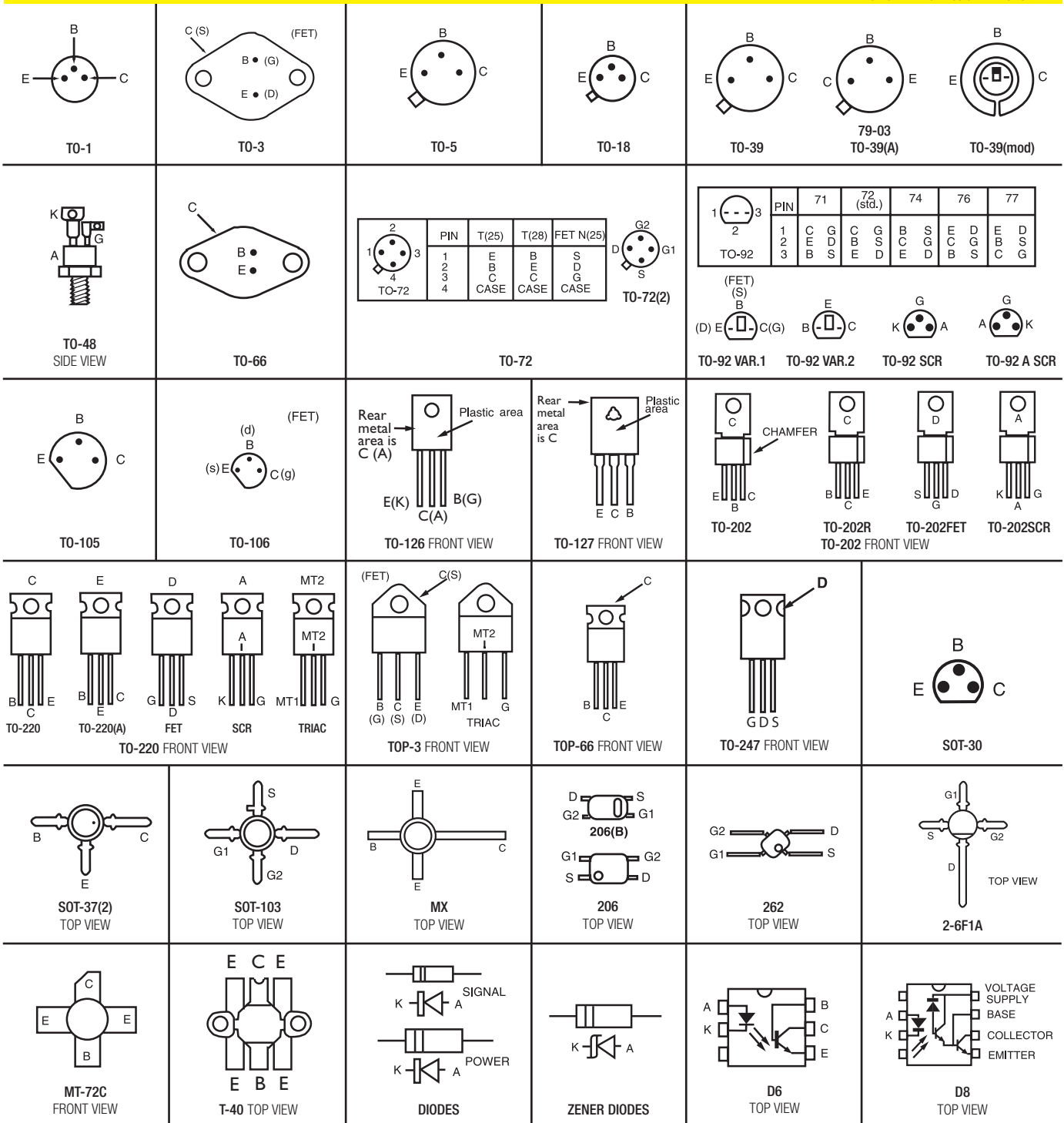
### Z 4710 BR100/03

Breakover voltage at $dV/dt=10V/ms$	$V_{B0}$	28 to 36V
Breakover voltage symmetry	$V_s$	3V
Breakover current at $V=0.98 V_{B0}$	$I_{B0}$	100µA
Maximum power dissipation	$P_{TOT}$	150mW



# SEMICONDUCTOR OUTLINES

BOTTOM VIEW UNLESS OTHERWISE STATED



## NTC THERMISTOR

These Negative Temperature Coefficient (NTC) Thermistors are useful for temperature sensing, temperature compensation, automatic gain control and fluid flow monitoring. They are not for use in conducting, reducing or aggressive gases or fluids. The resistance of the R1890 and R1895 can be calculated from the formula:

$$R = Ae^{B/T}$$

where, **R** is the thermistor resistance in ohms,  
**T** is the temperature in Kelvin,  
**e** is the natural logarithm base (2.718) and  
**A & B** are constants which should be calculated from measurements on each particular thermistor for best accuracy.

For nominal values of B and R, A is approximately 0.0277.

### SPECIFICATIONS

Maximum permissible current Operating temperature range  
 at zero power  
 Thermal time constant  
 Resistance value at 25°C  
 B-Constant (@ 25° and 85°C)

### R 1890

25mA  
 -20°C to +125°C  
 18s  
 100k ohms  
 4400

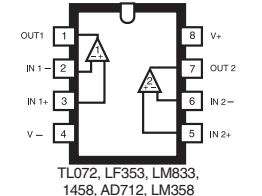
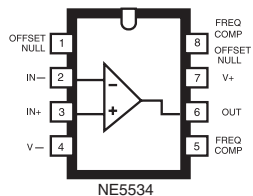
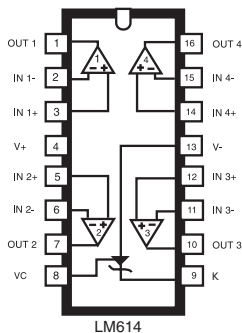
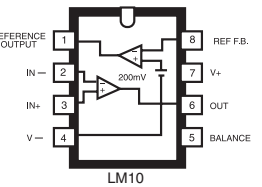
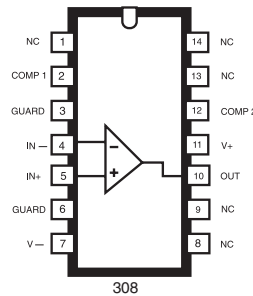
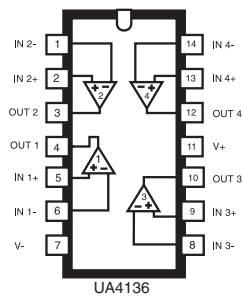
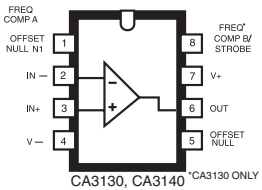
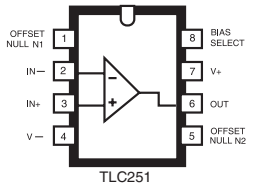
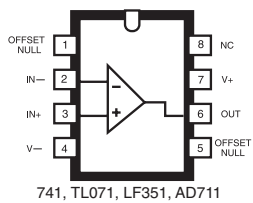
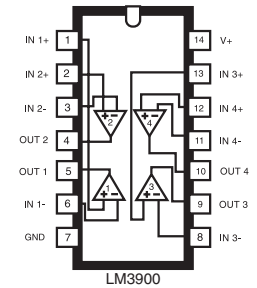
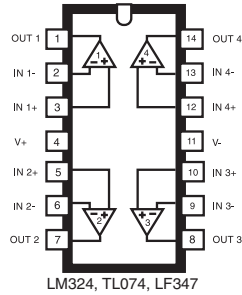
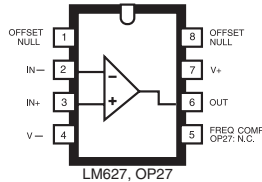
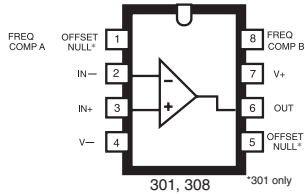
### R 1895

10mA  
 -30°C to +125°C  
 24s  
 100k ohms  
 4400

## OPERATIONAL AMPLIFIERS HANDY REFERENCE GUIDE

TYPE	DEVICE	INPUT OFFSET VOLT. MAX (mV)	INPUT OFFSET CURRENT MAX (nA)	INPUT BIAS CURRENT MAX (nA)	VOLT. GAIN MIN (VOLTS/V)	UNITY GAIN B.W. TYP (MHz)	SLEW RATE AV=1 TYP (V/us)	OUTPUT VOLT. SWING R1=10K (VS=+/-15) (V)	SUPPLY MIN (V)	VOLTAGE MAX (V)	CMRR MIN (dB)	DIFFERENTIAL INPUT VOLT. (V)	SUPPLY CURRENT TA=25C (NOTE 1) MAX (mA)	INTERNAL FREQ. COMP.
<b>SINGLE OP AMPS</b>														
LM301	GP amplifier	10	70	300	15k	1	0.5	+/-12	+/-3	+/-18	70	+/-30	3	No
LM741	GP amplifier	7.5	300	800	15k	1	0.5	+/-12	+/-3	+/-18	70	+/-30	2.8	Yes
LM308	Low input current	10	1.5	10	15k	1	0.3	+/-13	+/-2	+/-18	80	+/-1.0(Note2)	0.8	No
NE5534	Low noise	5	400	2000	15k	10	6	+/-12	+/-3	+/-22	70	+/-0.5(Note2)	8	No
OP27	Low noise precision	0.22	135	150	450k	5	1.7	+/-11	+/-4	+/-22	96	+/-0.7(Note2)	5.7	Yes
TL071	JFET low noise	13	2	7	15k	3	13	+/-12	+/-3.6	+/-18	70	+/-30	2.5	Yes
LF351	JFET low noise	13	4	8	15k	4	13	+/-12	+/-4	+/-18	70	+/-30	3.4	Yes
TLC251	Prog low power CMOS	10	0.3	0.6	7.5k	0.7	0.6	7.8(Vs=10V)	1.4	18	60	+/-18	0.22	Yes
CA3130	MOSFET input	15	0.03	0.05	50k	4	10	+/-6(Vs=+/-8)	+/-2.5	+/-8	70	+/-8	15	No
CA3140	MOSFET input	15	0.03	0.05	20k	4.5	9	+/-12	+/-2	+/-18	70	+/-8	6	Yes
LM10CL	Inc. voltage Ref.	5	3	40	3k	0.1	*	6.94(Vs=7v)	1.1	7	74	+/-7	0.57	Yes
LM3080	Transconductance	5	500	5000	5.4mmho	2@Ao1	50	+/-12	+/-2	+/-18	80	+/-5	typ 1.1	No
LM627	Precision Op-amp	0.1	25	25	4M	14	4.5	+/-14.0	+/-3.5	+/-18	120	25mA(Note2)	4.8	Yes
<b>DUAL OP AMPS</b>														
LM1458	GP amplifier	7.5	300	800	15k	1	0.5	+/-12	+/-3	+/-18	70	+/-30	5.6	Yes
LM358	Low power	9	150	500	15k	1	0.5	+/-12	+/-1.5	+/-16	65	32	2	Yes
LM833	Hi-Fi	5	200	1000	30k	15	7	+/-12	+/-5	+/-18	80	+/-30	8	Yes
TL072	JFET low noise	13	2	7	15k	3	13	+/-12	+/-3.6	+/-18	70	+/-30	5	Yes
LF353	JFET low noise	13	4	8	15k	4	13	+/-12	+/-6	+/-18	70	+/-30	6.5	Yes
LM13600	Transconductance	4	600	8000	5.4mmho	2@Ao1	50	+/-12	+/-2	+/-18	80	+/-5	typ 2.6	No
<b>QUAD OP AMPS</b>														
LM324	GP amplifier	9	150	500	15k	1	0.5	+/-13	+/-1.5	+/-16	65	32	3	Yes
UA4136	GP low noise	7.5	300	800	15k	3	1	+/-12	+/-2	+/-18	70	+/-30	11	Yes
TL074	JFET low noise	13	2	7	15k	3	13	+/-12	+/-3.6	+/-18	70	+/-30	10	Yes
LF347	JFET low noise	13	4	8	15k	4	13	+/-12	+/-4.5	+/-18	70	+/-30	11	Yes
LM614	Inc. volt Ref.	7	50	250	25k	>1	0.8	12.4	+3	+36	80	36	typ 0.8	Yes
LM3900	Norton amp	*	*	200	1.2k	2.5	20	10	+4	+32	*	*	10	Yes

\* Not Specified. Note 1: Supply current for all channels in the package. Note 2: Inputs have shunt diode protection, current must be limited. Note 3: Dick Smith Electronics may not stock all devices listed.



### AMPLIFIER CONFIGURATIONS

**Inverting Amplifier**

$$V_{out} = -R_2 / R_1 \times V_{in}$$

INPUT IMPEDANCE = R1

**Non-Inverting Amplifier**

$$V_{out} = V_{in} (R_1 + R_2) / R_1$$

**Voltage Follower**

$$V_{out} = V_{in}$$

# 555 TIMER (LM/NE555 BIPOLAR, TLC555 CMOS)

## GENERAL

The 555 Timer is a highly versatile low-cost IC specifically designed for precision timing applications. It can also be used in monostable multi-vibrator, astable multi-vibrator and Schmitt trigger applications.

## BIPOLAR 555

The 555 has many attractive features. It can operate from 4.5 volts to 16 volts. Its output can source (supply) or sink (absorb) load current up to a maximum of 200mA and so can directly drive loads such as relays, LEDs, low-power lamps, and high impedance speakers. When used in the 'timing' mode, the IC can readily produce accurately timing periods variable from a few microseconds to several hundred seconds via a single R-C network. Timing periods are virtually independent of supply rail voltage, have a temperature coefficient of only .005% per °C, can be started via a TRIGGER command signal, and can be aborted by a RESET command signal.

When used in the monostable mode, the IC produces output pulses with typical rise and fall times of a mere nS.

When used in the astable mode both the frequency and the duty cycle of the waveform can be accurately controlled with two external resistors and one capacitor.

## CMOS 555

Due to its high impedance inputs (typically  $10^{12}\Omega$ ), it is capable of producing accurate time delays and oscillations while using less expensive, smaller timing capacitors than NE555. Like the NE555, the TLC555 achieves both monostable (using one resistor and one capacitor) and astable (using two resistors and one capacitor) operation. In addition, 50% duty cycle astable operation is possible using only a single resistor and one capacitor. The TLC555 will operate at frequencies up to 2MHz and is fully compatible with CMOS, TTL, and MOS logic.

While the complementary CMOS output is capable of sinking over 100mA and sourcing over 10mA, the TLC555 exhibits greatly reduced supply current spikes during output transitions. This minimises the need for the large decoupling capacitors required by the NE555.

## MONOSTABLE OPERATION

In the monostable mode, the timer functions as a one-shot. Referring to the circuit the external capacitor is initially held discharged by a transistor inside the timer.

When a negative trigger pulse is applied to lead 2, the flip-flop is set, releasing the short circuit across the external capacitor, driving the output HIGH. The voltage across the capacitor, increases exponentially with the time constant  $t = R1C1$ . When the voltage across the capacitor equals  $2/3 VCC$ , the comparator resets the flip-flop which then discharges the capacitor rapidly and drives the output to its LOW state.

The circuit triggers on a negative-going input signal when the level reaches  $1/3 VCC$ . Once triggered, the circuit remains in this state until the set time has elapsed, even if it is triggered again during this interval. The duration of the output HIGH state is given by  $t = 1.1 R1C1$ . The timing interval is independent of supply. When Reset is not used it should be tied high to avoid any possibility of false triggering.

## ASTABLE OPERATION

When the circuit is connected as shown it triggers itself and free runs as a multivibrator. The external capacitor charges through  $R1$  and  $R2$  and discharges through  $R2$  only. Thus the duty cycle may be precisely set by the ratio of these two resistors.

In the astable mode of operation,  $C1$  charges and discharges between  $1/3 VCC$  and  $2/3 VCC$ . As in the triggered mode, the charge and discharge times and therefore frequency are independent of the supply voltage.

The charge time (output HIGH) is given by:

$$t1 = 0.693 (R1 + R2) C1$$

and the discharge time (output LOW) by:

$$t2 = 0.693 (R2) C1$$

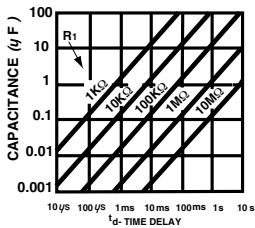
Thus the total period T is given by:

$$T = t1 + t2 = 0.693 (R1 + 2R2) C1$$

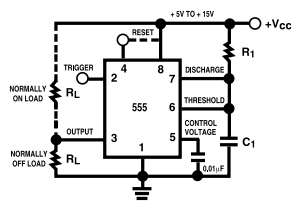
The frequency of oscillation is then:

$$f = \frac{1}{T} = \frac{1.44}{(R1 + 2R2) C1}$$

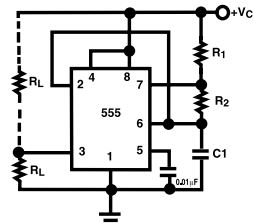
The duty cycle is given by:  $D = \frac{R2}{R1 + 2R2}$



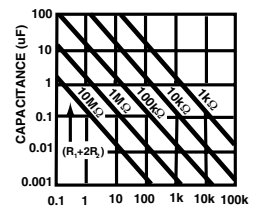
Time Delay vs R1 and C1



Basic Monostable Circuit



Basic Astable Circuit



Free Running Frequency

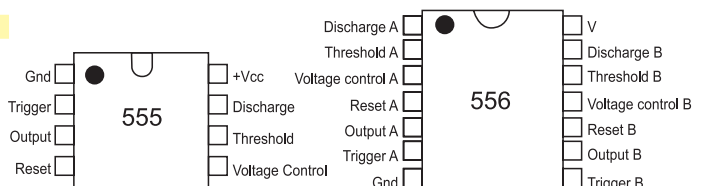
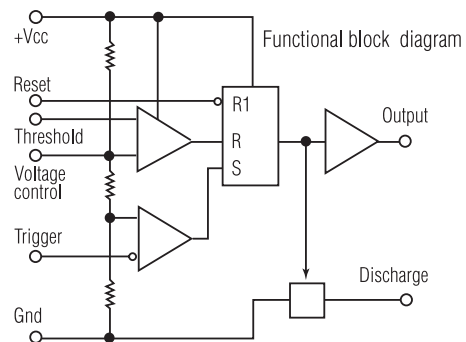
## SPECIFICATIONS

VCC = 15V UNLESS OTHERWISE STATED

TA = 25°C PARAMETERS	LM/NE555C LIMITS			TLC555C LIMITS		
	CONDITIONS	MIN	TYP MAX	CONDITIONS	MIN	TYP MAX
Supply Voltage		4.5	16V		2	18
Supply Current	(RL=∞, Vop=low)	10mA	15mA	(RL=∞, Vop=low)		360uA
Power Dissipation			600mW			600mW
Threshold Voltage	(% of Vcc)		67%	(% of Vcc)		67%
Threshold Current		0.1	0.25uA	(Vcc=5V)		10pA
Trigger Voltage	(% of Vcc)		33%	(% of Vcc)		33%
Trigger Current		0.5uA		(Vcc=5V)		10pA
Reset Voltage		0.4	0.5 1V		0.4	1.1 1.5
Reset Current		0.1	0.4mA	(Vcc=5V)		10pA
High Level Output	(IOL=100mA)	12.75	13.3V	(IOL=10mA)	12.5V	14.2V
Low Level Output	(IOL=100mA)	2	2.5V	(IOL=100mA)	1.28V	3.2V
Output Current	Sink		200mA			100mA
	Source		200mA			10mA
Initial Error of Timing Interval		1%			1%	

## FUNCTION TABLE

RESET	TRIGGER VOLTAGE	THRESHOLD VOLTAGE	OUTPUT	DISCHARGE SWITCH
Low	Irrelevant	Irrelevant	Low	On
High	<1/3 Vcc	Irrelevant	High	Off
High	>1/3 Vcc	>2/3 Vcc	Low	On
High	>1/3 Vcc	<2/3 Vcc	As previously established	



## VOLTAGE REGULATORS

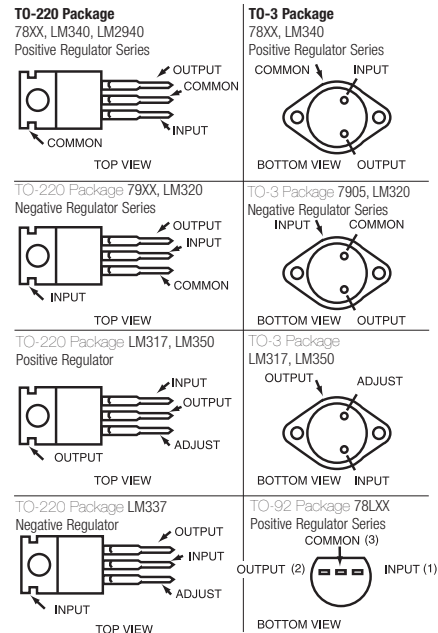
### THREE TERMINAL VOLTAGE REGULATORS

These voltage regulators almost make power supply building unnecessary since they require only a filtered DC voltage input. They are essentially indestructible (if used within manufacturer's specs) because of internal current limiting and thermal shutdown should a short occur. They are ideally suited to local, on-board regulation simplifying power supply distribution systems. Excellent for TTL and project supplies. With the advent of microprocessors and microcomputers, these regulators have been used extensively in power supplies for such systems. This type of use typifies their versatility and reliability. Another area of supply regulation use is with analogue operational amplifiers. These circuits usually call for both + and - complementary rail voltages. These three terminal devices ideally suit such requirements.

The addition of protection diodes, as shown in the 317/350 application circuit, is recommended if there is any possibility of the regulator input or output being shorted to ground. These may also be necessary if significant (1uF) capacitance is connected between ground and either the common or output terminals of the regulator.

TYPE	POL	CASE	VIN MAX	VIN MIN	VOUT TYP.	IOUT MAX (NOM)	DROP- OUT VOLT.	
78L05Z	pos.	TO-92	30	4.8	5	5.2	0.1	2.2
78L12Z	pos.	TO-92	30	11.5	12	12.5	0.1	2.2
78L15Z	pos.	TO-92	35	14.4	15	15.6	0.1	2.2
7805T	pos.	TO-220	35	4.8	5	5.2	1	2.5
7808T	pos.	TO-220	35	7.7	8	8.3	1	2.5
7812T	pos.	TO-220	35	11.5	12	12.5	1	2.5
7815T	pos.	TO-220	35	14.4	15	15.6	1	2.5
7805K	pos.	TO-3	35	4.8	5	5.2	1	2
7812K	pos.	TO-3	35	11.5	12	12.5	1	2
7815K	pos.	TO-3	35	14.4	15	15.6	1	2
7905T	neg.	TO-220	35	-4.8	-5	-5.2	1	2.3
7912T	neg.	TO-220	35	-11.5	-12	-12.5	1	2.3
7915T	neg.	TO-220	35	-14.4	-15	-15.6	1	2.3
7905K	neg.	TO-3	35	-4.8	-5	-5.2	1	2.3
LM317T	pos.	TO-220	40	adj.	1.2 to 37	1.5	3	
LM317K	pos.	TO-3	40	adj.	1.2 to 37	1.5	3	
LM350T	pos.	TO-220	35	adj.	1.2 to 32	3	3	
LM350K	pos.	TO-3	35	adj.	1.2 to 32	3	3	
LM337T	neg.	TO-220	40	adj.	-1.2 to -37.5	3	3	
LM-2940CT-5	pos.	TO-220	26	4.75	5	5.25	1	<1
LM-2940T-8	pos.	TO-220	26	7.6	8	8.4	1	<1
LM-2940T-10	pos.	TO-220	26	9.5	10	10.5	1	<1
LM-2940CT-12	pos.	TO-220	26	11.4	12	12.6	1	<1
LM-2941CT	pos.	TO-220	26V	adj.	5 to 20	1.5	<1	

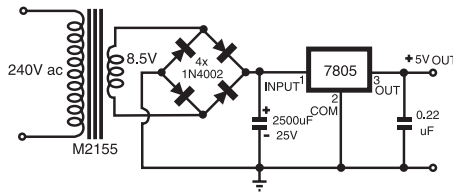
For all devices listed, the maximum junction temperature is 125°C. Except for TO-92 packages, maximum junction to case thermal resistance is 4°/watt. For TO-92 packages, junction to ambient thermal resistance is 230°/W.



## TYPICAL POWER SUPPLY

### OUTPUT: 5V @ 1A

The same basic circuit can be used with other regulators of different voltages, only the input AC voltage has to be changed to accommodate the requirement of the regulator, E.g. if an output of 12V was the requirement, a 7812 IC could be used with an AC input voltage of 15V.



## LM335 PRECISION TEMPERATURE SENSOR

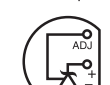
The LM335 operates as a two terminal zener having a breakdown voltage directly proportional to absolute temperature (10mV per Kelvin).

An adjustment terminal is provided, and maximum accuracy is realised when this is used to set the output voltage to 2.982 volts @ 25°C.

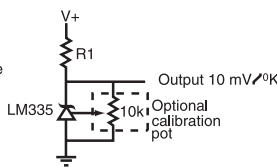
Max. rev. current	15mA
Continuous operating temp.	-40°C to 100°C
Operating output voltage (Tc = 25°C, I = 1mA) typ	2.98V
Max. uncalibrated temp. error (Tc = 25°C, I = 1mA)	6°C
Output voltage tempco.	10mV/°C
Dynamic impedance	0.6 ohms
Non-linearity (I = 1mA)	1.5°C



TO-46  
Metal can package



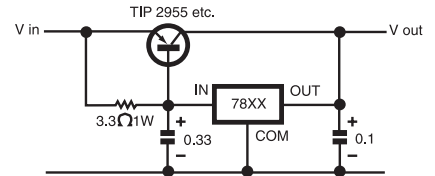
TO-92  
Plastic package



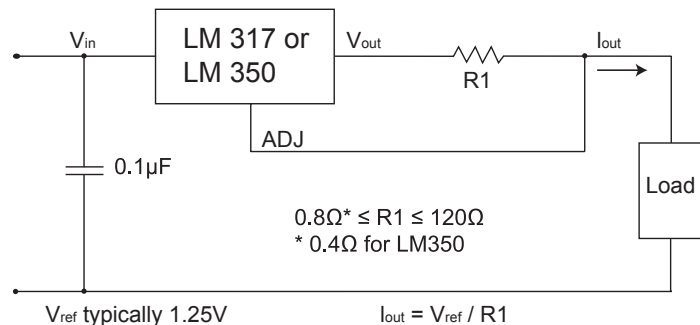
## INCREASING THE OUTPUT CURRENT OF THE REGULATOR

By adding a PNP power transistor to a positive regulator, the output current can be increased above the normal rating of the regulator itself. The circuit shown below can be expected to deliver in excess of 4A with the pass transistor mounted on a heatsink. The same conditions can also be implemented with a negative regulator, the difference being the polarities of components, the pass transistor in this case would be an NPN type such as a 2N3055.

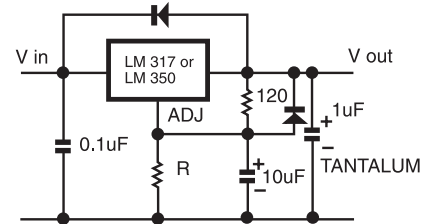
### CURRENT BOOSTING



## USING THE LM317 OR LM350 AS A CURRENT REGULATOR



## LM317 OR LM350 SERIES



- Choose Resistor R as follows:  
 $R = (96 \times V_{out}) - 120$   
Where R is in ohms and Vout is in volts.
- Vin should be at least 2.5 volts greater than Vout.
- Capacitor voltage ratings must be chosen appropriately.
- The protection diodes shown will be necessary if the input or output of the regulator is shorted to ground. 1A types should be adequate.

## LM334 ADJUSTABLE CURRENT SOURCE

Operating voltage

1-30

Max power dissipation

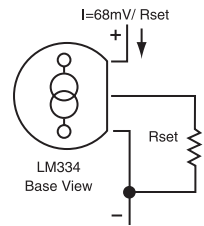
200mW

Voltage sensitivity

0.1%/V

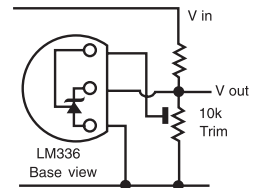
Temperature sensitivity

+0.33%/°C



## LM336 REFERENCE DIODES

	LM336-2.5	LM336-5.0
Max fwd current (mA)	10	10
Untrimmed ref voltage	2.44-2.54	4.9-5.1
Dynamic impedance (ohms)	1	2
Temp stability (0-70°C)	20ppm	20ppm

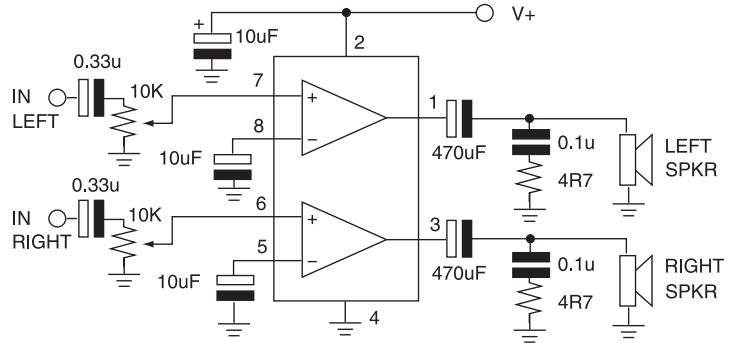


## Z 6070 TDA2822M DUAL LOW-VOLTAGE POWER AMPLIFIER

The TDA2822M offers low voltage, low quiescent current operation and high power output from an 8-pin DIP package.

### SPECIFICATIONS:

<b>Supply voltage:</b>	1.8-15 volt
<b>Quiescent current:</b>	9mA max
<b>Gain:</b>	39dB typ
<b>Channel balance:</b>	1dB
<b>Input resistance:</b>	100k ohm
<b>Power output (typ.):</b>	220mW @ 6 volt, 16 ohms
<b>(1kHz, 10% dist.):</b>	380mW @ 6 volt, 8 ohms
	1000mW @ 9 volt, 8 ohms
	650mW @ 6 volt, 4 ohms
	110mW @ 3 volt, 4 ohms
<b>Rth j-amb:</b>	100 K/W
<b>Rth j-pin:</b>	70 K/W
<b>Max junction temp:</b>	150°C



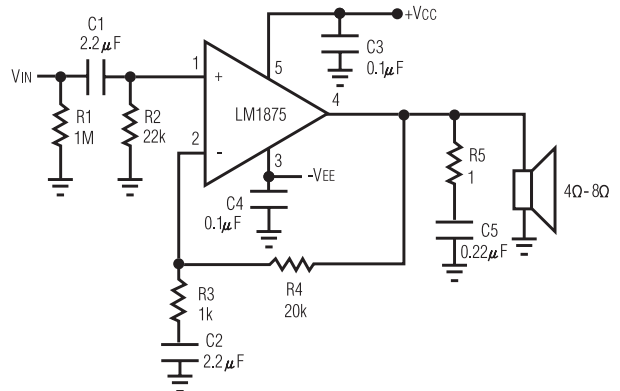
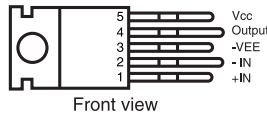
## LM1875 20 WATT AMPLIFIER IC

The LM1875 is a monolithic power amplifier IC which takes advantage of advanced techniques to achieve low distortion, even at high output levels. Other features include high gain, fast slew rate and a wide power bandwidth, large output voltage swing, high current capability and wide supply range. The device is internally compensated for gains of 10 and greater.

### SPECIFICATIONS:

<b>Supply voltage:</b>	+/-30 volt max
<b>Quiescent current:</b>	100mA max
<b>THD 20W/1KHz:</b>	0.015% typ
<b>Open loop gain:</b>	90dB typ
<b>Current limit:</b>	4A typ
<b>Thermal resistance:</b>	2°C/W typ
<b>(junction to case)</b>	
<b>Junction temperature:</b>	150°C max

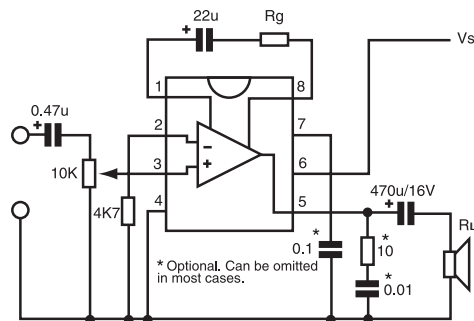
TO-220 Power package (T)



## LM386 POWER AMPLIFIER

The LM386 power amplifier IC is ideal for low voltage applications. The gain is internally set at 20, but may be increased up to 200 by adding components between pins 1 and 8.

Gain	Rg	Output Power (mW)			
		Vs (volts)	4	8	16
20	∞	12	380	660	780
50	680	9	380	560	400
100	180	6	250	250	150
200	0	5	190	160	90



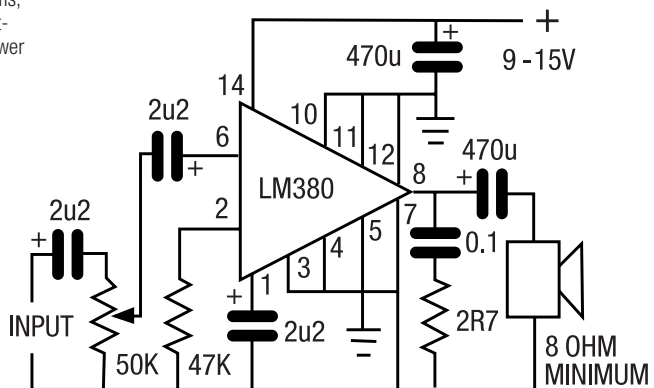
## LM380 2.5 WATT AMPLIFIER IC

The LM380 is an amplifier IC for general purpose audio applications such as intercoms, small audio amplifiers, headphone amplifiers, TV and radio sound. The output is short-circuit proof and the device has internal thermal limiting. It is also suitable for low power DC operation such as for controlling a small motor or servo control.

The gain is internally set at 34dB, a voltage gain of 50.

### SPECIFICATIONS:

<b>Supply voltage:</b>	10-22 volts
<b>Gain:</b>	34dB
<b>Output power:</b>	2.5 watt (min.)
	(into 8 ohms, @ 3% THD)
<b>Peak current:</b>	1.3 amps
<b>Input resistance:</b>	150k ohms
<b>Bandwidth:</b>	100kHz
<b>THD:</b>	<0.2%
<b>Quiescent current:</b>	7mA typ



## COMMON CONNECTORS

### DIN CONNECTORS

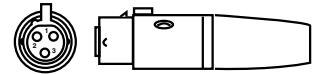
Contact Arrangement	Application		Connections				
			1	2	3	4	5
MALE PLUG	Microphone	Mono (balanced)	Hot lead	Screen earth	Return lead		
		Mono (unbalanced)	Hot lead				
	Microphone	Stereo (balanced)	Hot lead of left-hand channel		Return lead of left-hand channel	Hot lead of right-hand channel	Return lead of right-hand channel
Stereo (unbalanced)		Hot lead of left-hand channel			Hot lead of right-hand channel		
FEMALE SOCKET	Turntable	Monaural system			Hot lead		Connected to 3
		Stereophonic system		Hot lead of left-hand channel		Hot lead of right-hand channel	
	Tape Recorder	Monaural system	Input signal	Output signal		Connected to 1	Connected to 3
Stereophonic system		Input signal of left-hand channel	Output signal of left-hand channel	Input signal of right-hand channel	Output signal of right-hand channel		

### 3 PIN XLR CONNECTORS

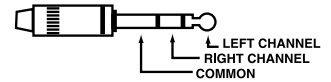
Widely used in professional audio applications these connectors generally offer a locking mechanism, very good cable clamping and low contact resistance. The pins are arranged so that when the connectors mate, one pin (used as the ground pin) always makes contact before the other two.

#### IEC STANDARD COMMON ALTERNATIVE

- |                        |                 |
|------------------------|-----------------|
| 1. Earth/Shield        | 1. Earth/Shield |
| 2. In Phase (hot)      | 2. Out of Phase |
| 3. Out of Phase (cold) | 3. In Phase     |

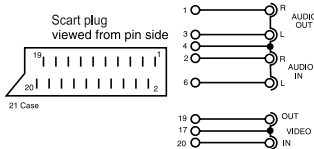


### HEADPHONE PLUG



### SCART CONNECTOR

The Scart connector, also known as the Euroconnector or Peri-Television connector, is part of a system for connecting television receivers and other home entertainment equipment. A Scart connector has 21 pins which provide for stereo audio and composite video in and out, RGB, two data lines and two control lines. A variation allows for separate chrominance and luminance signals.

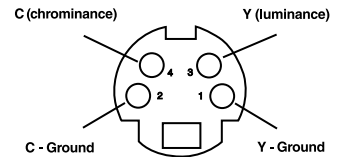


#### NOTES:

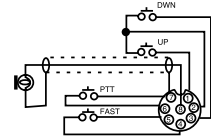
- Pin 8 is used to select between an internal and external composite video signal. +12 volts applied to this pin will enable the composite video input on pin 20. Alternatively, this input may be enabled via a control on the video display device, e.g. an external AV select switch.
- Pin 16 is a control line to select the external RGB inputs. +3 volts applied to this pin enables the RGB inputs on pins 15, 11 and 7.
- When used for S-VHS signals, pin 15 carries chrominance and pin 20 luminance.
- Pin 10 and 14 have been assigned for data for controlling and monitoring other appliances.

PIN	USE	LEVEL/IMPEDANCE
1	Audio Output (R)	0.5V/1kΩ
2	Audio Input (R)	0.5V/10kΩ
3	Audio Output (L)	0.5V/1kΩ
4	Audio Ground	
5	Blue Ground	
6	Audio Input (L)	0.5V/10kΩ
7	Blue	0.7Vp-p/75Ω
8	Status (CVBS)	L: 0-2V H: 10-12V/10kΩ
9	Green Ground	
10	Data D2B (Inverted)	
11	Green	0.7Vp-p/75Ω
12	Data D2B	
13	Red Ground	
14	D2B Ground	
15	Red	0.7Vp-p/75Ω
16	RGB Status/Fast Blanking	L: 0-0.4V H: 1-3V/75Ω
17	CVBS Video Ground	
18	RGB Status Ground	
19	Composite Video Output	1V/75Ω
20	Composite Video Input	1V/75Ω
21	Case/shield	

### S-VIDEO CONNECTOR



### YAESU 8-PIN MICROPHONE

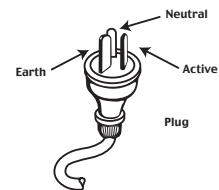


Commonly used on base and mobile transceivers

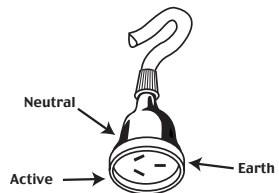
### GET IT RIGHT!

Note the locations of the Active, Neutral and Earth pins, and their associated wire colours in the table below. If you are replacing a mains plug or socket on an extension cord, then follow exactly the wiring instructions supplied with the mains plug or socket.

### PLUG



### SOCKET

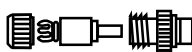


#### COLOUR CODE TABLE

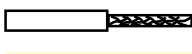
CONDUCTOR	COLOUR	OLD COLOUR
Active	Brown	Red
Neutral	Blue	Black
Earth	Green/Yellow	Green

### WIRING RF PLUGS

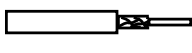
#### BELLING-LEE TYPE



Step 1: Cut end of cable evenly and remove 22mm of outer sheath.



Step 2: Remove 16mm of braid and dielectric, taking care not to damage centre conductor. Lightly tin conductor.



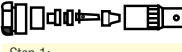
Step 3: Slide cap and collet clamp over cable and slightly squeeze collet in position. Comb braid out and fan back over collet.



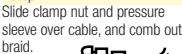
Step 4: Solder tip of plug to centre conductor. Do not overheat. Push sub-assembly into body of plug, and screw cap onto body to complete assembly.



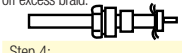
#### BNC TYPE (FEMALE)



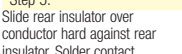
Step 1: Cut end of cable evenly and remove 8mm of outer sheath.



Step 2: Slide clamp nut and pressure sleeve over cable, and comb out braid.



Step 3: Fold braid back and insert ferrule between braid and dielectric. Trim off excess braid.



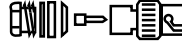
Step 4: Remove 5mm of dielectric without damaging centre conductor. Tin conductor.

Step 5: Slide rear insulator over conductor hard against rear insulator. Solder contact to conductor.

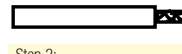
Step 6: Fit front insulator in body and push sub-assembly onto body. Slide pressure sleeve into body and screw in the clamp nut to clamp cable.



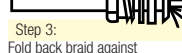
#### BNC TYPE (MALE)



Step 1: Cut end of cable evenly, and remove 7mm of outer sheath.



Step 2: Slide clamp-nut, washer and rubber spacer over cable. Slide pressure sleeve over braid, and fan out braid.



Step 3: Fold back braid against pressure sleeve and trim off excess. Strip 4mm of inside insulation off centre wire. Solder pin onto centre wire.



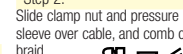
Step 4: Fit assembly, pin first, into main receptacle. Push washer, rubber spacer and clamp nut up and screw into place.



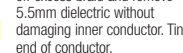
#### 'N' TYPE



Step 1: Cut end of cable evenly and remove 9mm of outer sheath.



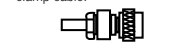
Step 2: Slide clamp nut and pressure sleeve over cable, and comb out braid.



Step 3: Fold braid back and insert ferrule between braid and dielectric. Trim off excess braid and remove 5.5mm dielectric without damaging inner conductor. Tin end of conductor.



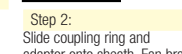
Step 4: Slide rear insulator over conductor and position against end of dielectric. Slide contact over conductor until hard against rear insulator. Solder contact to conductor.



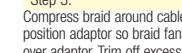
#### UHF TYPE



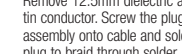
Step 1: Cut end of cable evenly and remove 27mm of outer sheath.



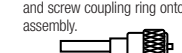
Step 2: Slide coupling ring and adaptor onto sheath. Fan braid back over sheath.



Step 3: Compress braid around cable and position adaptor so braid fans over adaptor. Trim off excess braid.



Step 4: Remove 12.5mm dielectric and tin conductor. Screw the plug assembly onto cable and solder plug to braid through solder holes.





## THE RS-232C INTERFACE

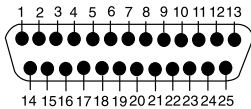
### THE BASIC RS-232C ELECTRICAL SPECIFICATION

- Communication rate:  
0-20,000 bits per second
- Driver output voltage levels maximum no load:  
-25V logic 1, +25V logic 0
- Driver output voltage ranges for loads between 3k and 7kΩ  
logic 1: -15V (7k) and -5V (3k)  
logic 0: +15V (7k) and +5V (3k)
- Driver output current, short circuit:  
500mA maximum
- Driver output impedance with power off:  
300Ω minimum
- Maximum driver output slew rate:  
30V per microsecond
- Receiver input resistance:  
7kΩ maximum  
3kΩ minimum
- Effective receiver input capacitance:  
2500pF maximum
- Maximum receiver input voltage range:  
-25V to +25V

The following table shows the nine most common signals used for serial communications and their pin allocations for nine and 25 pin connectors.

NAME	SIGNAL (9 PIN)	PIN NO. (25 PIN)	PIN NO.
TD	Transmitted Data	3	2
RD	Received Data	2	3
RTS	Request to Send	7	4
CTS	Clear to Send	8	5
DSR	Data Set Ready	6	6
SG	Signal Ground	5	7
CD	Carrier Detect	1	8
DTR	Data Terminal Ready	4	20
RI	Ring Indicator	9	22

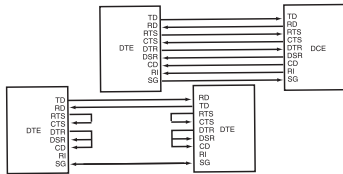
### DB25 SERIAL MALE PLUG (FRONT VIEW)



### COMMON RS-232C CONNECTIONS

Each signal (except ground) is either an input or an output depending whether it is on the Data Terminal Equipment 'DTE' (e.g. a computer terminal) or the Data Circuit Terminating Equipment 'DCE' (e.g. a modem). A typical connection is shown here.

When two computer terminals (DTEs) are connected together a 'null modem' hook-up is needed to fool each DTE into thinking that it is connected to DCE. A typical null modem connection is shown here.



## ASCII CHARACTER CODES AS DEFINED IN ANSI X3.4

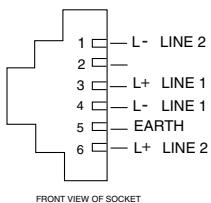
DEC	HEX	CHAR	DEC	HEX	CHAR	DEC	HEX	CHAR	DEC	HEX	CHAR
0	0	NUL	33	21	!	66	42	B	99	63	c
1	1	SOH	34	22	"	67	43	C	100	64	d
2	2	STX	35	23	#	68	44	D	101	65	e
3	3	ETX	36	24	\$	69	45	E	102	66	f
4	4	EOT	37	25	%	70	46	F	103	67	g
5	5	ENQ	38	26	&	71	47	G	104	68	h
6	6	ACK	39	27	'	72	48	H	105	69	i
7	7	BEL	40	28	(	73	49	I	106	6A	j
8	8	BS	41	29	)	74	4A	J	107	6B	k
9	9	HT	42	2A	*	75	4B	K	108	6C	l
10	A	LF	43	2B	+	76	4C	L	109	6D	m
11	B	VT	44	2C	,	77	4D	M	110	6E	n
12	C	FF	45	2D	-	78	4E	N	111	6F	o
13	D	CR	46	2E	.	79	4F	O	112	70	p
14	E	SO	47	2F	/	80	50	P	113	71	q
15	F	SI	48	30	0	81	51	Q	114	72	r
16	10	DLE	49	31	1	82	52	R	115	73	s
17	11	DC1	50	32	2	83	53	S	116	74	t
18	12	DC2	51	33	3	84	54	T	117	75	u
19	13	DC3	52	34	4	85	55	U	118	76	v
20	14	DC4	53	35	5	86	56	V	119	77	w
21	15	NAK	54	36	6	87	57	W	120	78	x
22	16	SYN	55	37	7	88	58	X	121	79	y
23	17	ETB	56	38	8	89	59	Y	122	7A	z
24	18	CAN	57	39	9	90	5A	Z	123	7B	{
25	19	EM	58	3A	:	91	5B	[	124	7C	
26	1A	SUB	59	3B	;	92	5C	\	125	7D	}
27	1B	ESC	60	3C	<	93	5D	]	126	7E	~
28	1C	FS	61	3D	=	94	5E	^	127	7F	DEL
29	1D	GS	62	3E	>	95	5F	-			
30	1E	RS	63	3F	?	96	60	'			
31	1F	US	64	40	@	97	61	a			
32	20	SPACE	65	41	A	98	62	b			

## ASCII CONTROL CODES

NUL	(Ctrl @)	Null	DLE	(Ctrl P)	Data Link Escape
SOH	(Ctrl A)	Start of Heading	DC1	(Ctrl Q)	Device Control 1
STX	(Ctrl B)	Start of Text	DC2	(Ctrl R)	Device Control 2
ETX	(Ctrl C)	End of Text	DC3	(Ctrl S)	Device Control 3
EOT	(Ctrl D)	End of Transmission	DC4	(Ctrl T)	Device Control 4
ENQ	(Ctrl E)	Enquiry	NAK	(Ctrl U)	Negative Acknowledge
ACK	(Ctrl F)	Acknowledge	SYN	(Ctrl V)	Synchronous Idle
BEL	(Ctrl G)	Bell	ETB	(Ctrl W)	End of Transmission Block
BS	(Ctrl H)	Backspace	CAN	(Ctrl X)	Cancel
HT	(Ctrl I)	Horizontal Tab	EM	(Ctrl Y)	End of Medium
LF	(Ctrl J)	Line Feed	SUB	(Ctrl Z)	Substitute
VT	(Ctrl K)	Vertical Tab	ESC	(Ctrl [)	Escape
FF	(Ctrl L)	Form Feed	FS	(Ctrl \)	File Separator
CR	(Ctrl M)	Carriage Return	GS	(Ctrl ])	Group Separator
SO	(Ctrl N)	Shift Out	RS	(Ctrl ^)	Record Separator
SI (Ctrl O)	Shift In		US	(Ctrl _)	Unit Separator

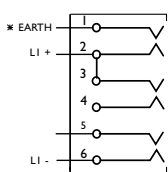
## COMMON TELEPHONE CONNECTIONS

### FCC SERIES



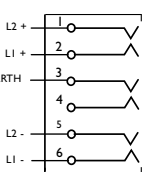
FRONT VIEW OF SOCKET

SIX POSITION SOCKET

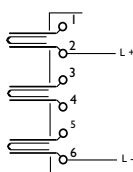


SINGLE LINE SOCKET

### 600 SERIES



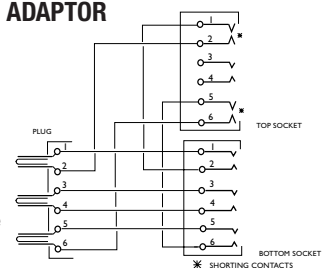
TWO LINE SOCKET



SINGLE LINE PLUG

### F 5114 MODE 3 DOUBLE ADAPTOR

The Mode 3 Double Adaptor is used with a fax machine or modem in the top socket, and an ordinary telephone in the bottom socket. The fax or modem, which must feature genuine 4-wire Mode 3 'change-over' line connection, can then switch the telephone in or out as part of its operation. Note the 'shorting contacts' in the top socket. These will allow the telephone to continue to work should the plug for the fax or modem be removed.



Front view of Socket

